

DESCRIPTION

The MP6550 is an H-bridge motor driver. The H-bridge consists of four N-channel power MOSFETs and an internal charge pump to generate gate-drive voltages. The MP6550 is typically used to drive a DC brush motor.

The MP6550 operates on a motor power supply voltage from 1.8V to 22V, which can supply an output current of up to 2A according to the logic control.

The MP6550 has a PWM input interface, which is compatible with industry-standard devices. Very low standby circuit current can be achieved when the device is disabled.

An internal current-sensing circuit provides an output with a voltage proportional to the load current. Cycle-by-cycle current regulation and limiting is also provided. These features do not require the use of a low-value shunt resistor.

There are internal shutdown functions for over-current protection, short-circuit protection, under-voltage lockout, and over-temperature protection.

The MP6550 requires a minimal number of readily available, standard external components, and is available in a QFN-12 (2mmx2mm) package.

FEATURES

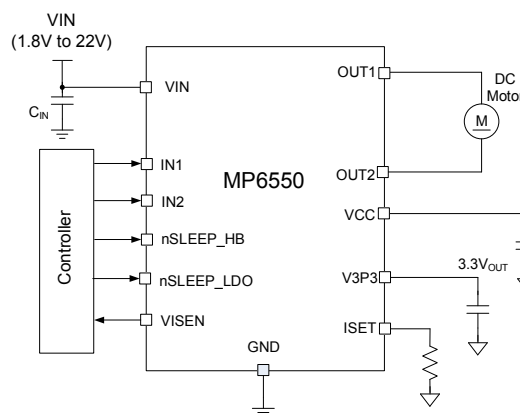
- Wide 1.8V to 22V Operating Input Range
- 2A Continuous Driver Current
- MOSFET On Resistance (HS + LS) 240mΩ
- Cycle-by-Cycle Current Regulation/Limit
- Built-In 3.3V Reference Output
- PWM Input Interface, Compatible with Industry-Standard Devices, Up to 100kHz
- Low Standby Circuit Current
- Thermal Shutdown
- Internal Charge Pump
- Cycle-by-Cycle Over-Current Protection
- Short-Circuit Protection
- Available in a QFN-12 (2mmx2mm) Package

APPLICATIONS

- Robotics
- Cameras
- Toys
- Consumer Products
- Medical Devices

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TYPICAL APPLICATION

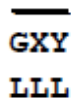


ORDERING INFORMATION

| Part Number* | Package | Top Marking |
|--------------|------------------|-------------|
| MP6550GG | QFN-12 (2mmx2mm) | See Below |

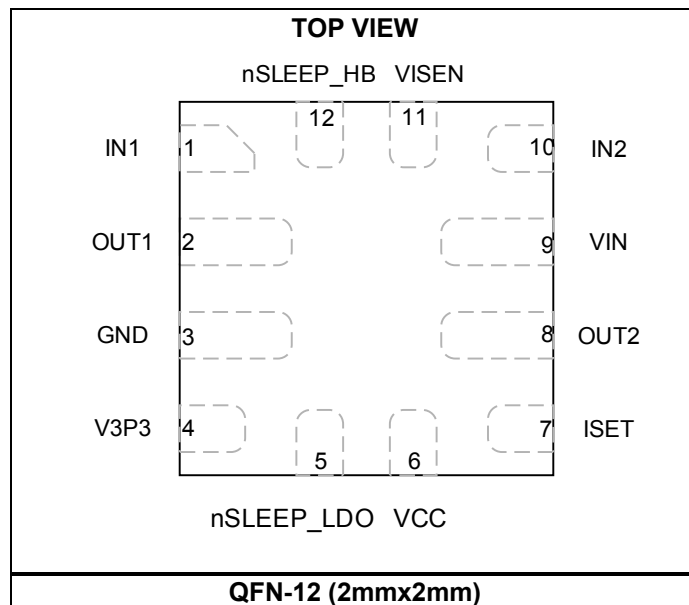
* For Tape & Reel, add suffix -Z (e.g. MP6550GG-Z).

TOP MARKING


GXY
LLL

GX: Product code of MP6550GG
 Y: Year code
 LLL: Lot number

PACKAGE REFERENCE



PIN FUNCTIONS

| Pin # | Name | Description |
|-------|------------|---|
| 1 | IN1 | Input 1. Internal pull-down resistor. |
| 2 | OUT1 | Switch output 1. Connect OUT1 to the motor winding. |
| 3 | GND | Ground. |
| 4 | V3P3 | 3.3V regulator output. Bypass to GND with a 0.47 μ F capacitor. |
| 5 | nSLEEP_LDO | Sleep mode input of LDO output. Pull high for 3.3V LDO normal operation. Pull low to disable the LDO output and enter low-power sleep mode. Internal pull-down. |
| 6 | VCC | Internal circuitry supply bypass. Bypass VCC to GND with a 0.1 μ F to 10 μ F capacitor. |
| 7 | ISET | Current programming resistor. Connect a resistor to ground to set the current limit and VISEN output voltage. |
| 8 | OUT2 | Switch output 2. Connect OUT2 to the motor winding. |
| 9 | VIN | Supply voltage. An input capacitor is needed to prevent large voltage spikes from appearing at the input. |
| 10 | IN2 | Input 2. Internal pull-down resistor. |
| 11 | VISEN | Current-sense output voltage. |
| 12 | nSLEEP_HB | Sleep mode input of H-bridge. Pull high for H-bridge normal operation. Pull low to disable the H-bridge output and enter low-power sleep mode. Internal pull-down. |

ABSOLUTE MAXIMUM RATINGS ⁽¹⁾

| | |
|--|-----------------------------------|
| Supply voltage (V_{IN}) | -0.3V to +24V |
| V_{OUTx} | -0.3V to $V_{IN} + 0.3V$ |
| All other pins | -0.3 to +5V |
| Continuous power dissipation ($T_A = 25^\circ C$) ⁽²⁾ | 1.56W |
| Junction temperature | 150 $^\circ C$ |
| Lead temperature | 260 $^\circ C$ |
| Storage temperature | -65 $^\circ C$ to +150 $^\circ C$ |

Recommended Operating Conditions ⁽³⁾

| | |
|-----------------------------------|-----------------------------------|
| Supply voltage (V_{IN}) | 1.8V to 22V |
| Operating junction temp (T_J) | -40 $^\circ C$ to +125 $^\circ C$ |

| | | |
|--|---------------|---------------|
| Thermal Resistance ⁽⁴⁾ | θ_{JA} | θ_{JC} |
| QFN-12 (2mmx2mm) | 80 | 16 |
| | $^\circ C/W$ | |

Notes:

- Exceeding these ratings may damage the device.
- The maximum allowable power dissipation is a function of the maximum junction temperature T_J (MAX), the junction-to-ambient thermal resistance θ_{JA} , and the ambient temperature T_A . The maximum allowable continuous power dissipation at any ambient temperature is calculated by P_D (MAX) = $(T_J$ (MAX) - T_A) / θ_{JA} . Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- Continuous current depends on PCB layout and ambient temperature.
- Measured on JESD51-7, 4-layer PCB.

ELECTRICAL CHARACTERISTICS

$V_{IN} = 5V$, $T_A = 25^\circ C$, unless otherwise noted.

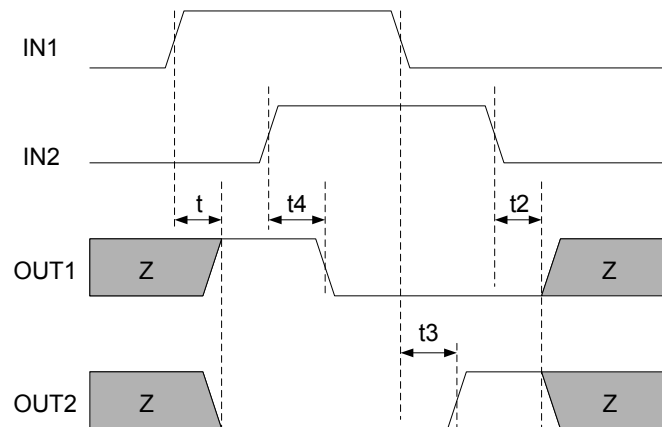
| Parameters | Symbol | Condition | Min | Typ | Max | Units |
|--|---------------|--|-----|------|------|------------|
| Power Supply | | | | | | |
| Input supply voltage | V_{IN} | | 1.8 | | 22 | V |
| Operating supply current | | no PWM, no load, nSLEEP_HB = 5V, nSLEEP_LDO = 0 | | 0.85 | 1.1 | mA |
| | | 50-kHz PWM, no load, nSLEEP_HB = 5V, nSLEEP_LDO = 0 | | 1.2 | 1.5 | mA |
| Operating supply current | | $V_{IN} = 5V$ to 22V, $I_{V3P3} = 0A$, nSLEEP_HB = 0, nSLEEP_LDO = 5V | | 80 | 120 | μA |
| Sleep mode supply current | I_Q | nSLEEP_HB = 0, nSLEEP_LDO = 0 | | 10 | 120 | nA |
| Under-voltage lockout threshold rising | | | | 1.68 | 1.8 | V |
| Under-voltage lockout threshold hysteresis | | | | 150 | | mV |
| Externally applied PWM frequency | f_{PWM} | | | | 100 | kHz |
| V3P3 Regulator | | | | | | |
| LDO output | | $V_{IN} > 3.7V$, $I_{OUT} = 0A$ to 50mA | 3.1 | 3.3 | 3.5 | V |
| Control Inputs (IN1, IN2, nSLEEP_HB, nSLEEP_LDO) | | | | | | |
| Input high voltage | V_{IH} | | 1.6 | | | V |
| Input low voltage | V_{IL} | | | | 0.4 | V |
| Input high current | I_{IH} | $V_{IN} = 3.3V$ | | | 50 | μA |
| Input low current | I_{IL} | $V_{IN} = 0V$ | -5 | | +5 | μA |
| Input pull-down resistance | RPD | | | 520 | | k Ω |
| Motor Driver Outputs (OUT1, OUT2) | | | | | | |
| HS switch on resistance | $R_{DS(ON)}$ | $I_O = 800mA$, $T_A = 25^\circ C$ | | 0.12 | 0.17 | Ω |
| LS switch on resistance | $R_{DS(ON)}$ | $I_O = 800mA$, $T_A = 25^\circ C$ | | 0.12 | 0.17 | Ω |
| Output enable time | t_1 | | | | 200 | ns |
| Output disable time | t_2 | | | | 200 | ns |
| Delay time | t_3 | INx high to OUTx high | | | 300 | ns |
| | t_4 | INx low to OUTx low | | | 300 | ns |
| Output rise time | | $R_L = 20\Omega$ | | | 100 | ns |
| Output fall time | | $R_L = 20\Omega$ | | | 100 | ns |
| Wake time, nSLEEP_HB rising edge to part H-bridge active | t_{WAKE} | | | | 200 | μs |
| Dead time | | | | 100 | | ns |
| Current Control | | | | | | |
| Off time | t_{ITRIP} | After ITRIP | | 17 | | μs |
| ISET current | I_{SET} | | 95 | 100 | 105 | $\mu A/A$ |
| Current trip voltage (rising) | $V_{ITRIP-R}$ | At VISEN pin | 475 | 500 | 525 | mV |
| Current trip voltage (falling) | $V_{ITRIP-F}$ | At VISEN pin | 380 | 400 | 420 | mV |

ELECTRICAL CHARACTERISTICS (continued)
 $V_{IN} = 5V$, $T_A = 25^\circ C$, unless otherwise noted.

| Parameters | Symbol | Condition | Min | Typ | Max | Units |
|--|--------------------|-------------------|-----|-----|-----|------------|
| VISEN Output | | | | | | |
| Output voltage accuracy | ΔV_{VISEN} | $V_{ISET} > 0.1V$ | -5 | | +5 | % |
| Protection Circuits | | | | | | |
| Current limit | I_{OCP} | | 4 | 5 | 6 | A |
| OCP retry time | t_{OCR} | | | 2.5 | | ms |
| Thermal shutdown threshold ⁽⁶⁾ | T_{TSD} | | | 160 | | $^\circ C$ |
| Thermal shutdown hysteresis ⁽⁶⁾ | | | | 25 | | $^\circ C$ |

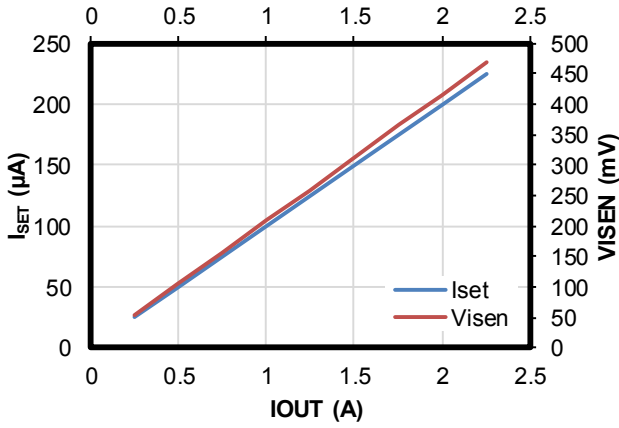
Note:

6) Guaranteed by design.


Figure 1: Input/Output Timing

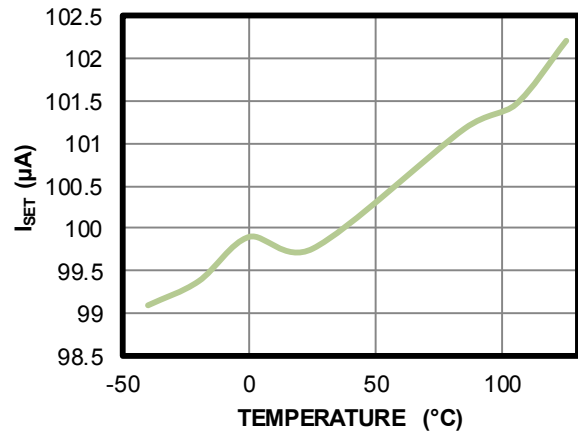
TYPICAL CHARACTERISTICS

Current Sense ($I_{SET} = 2k\Omega$)

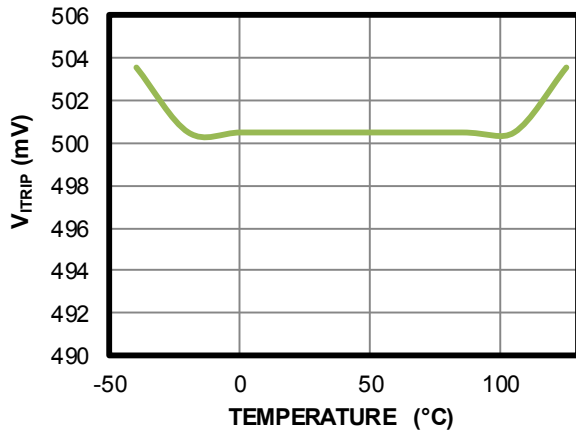


I_{SET} vs. Temperature

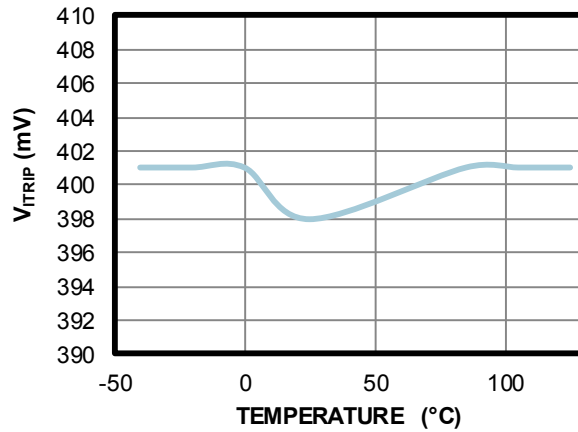
$I_{OUT} = 1A$



Current Trip Voltage (Rising) vs. Temperature

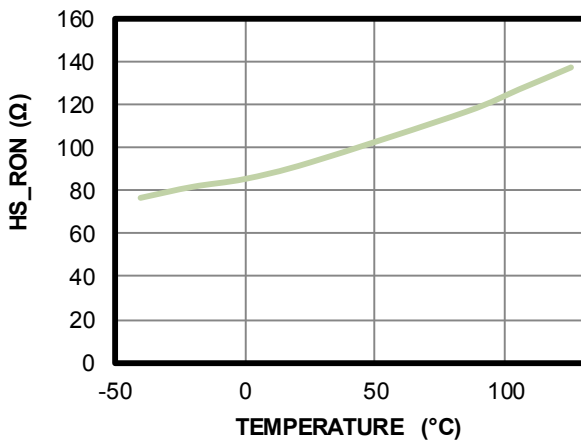


Current Trip Voltage (Falling) vs. Temperature



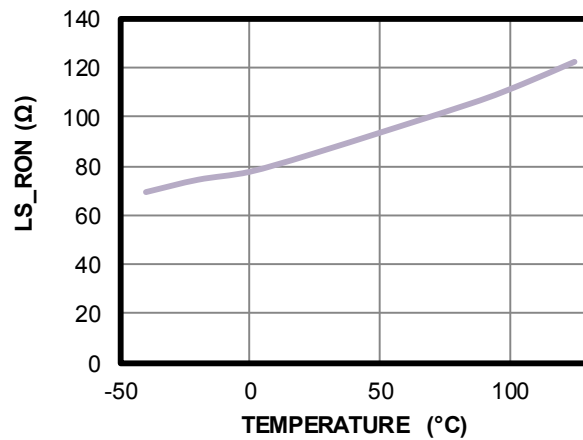
HS R_{ON} vs. Temperature

$I_{OUT} = 800mA$

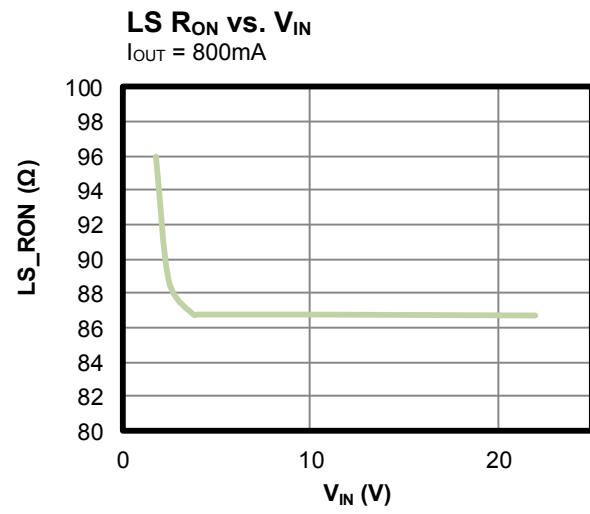
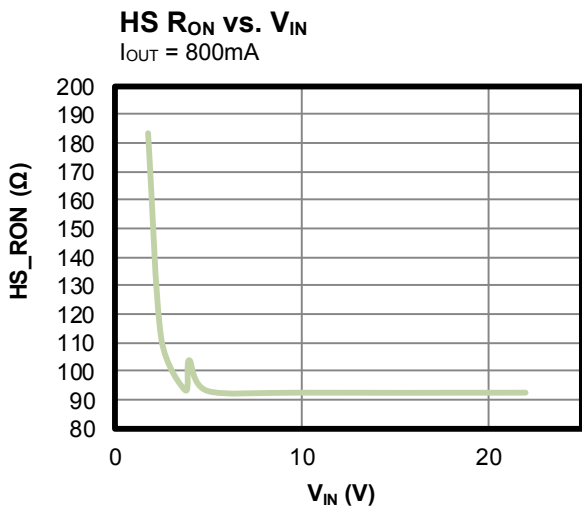


LS R_{ON} vs. Temperature

$I_{OUT} = 800mA$



TYPICAL CHARACTERISTICS (continued)

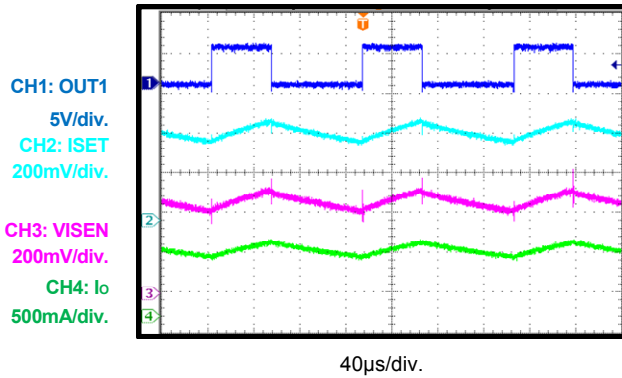


TYPICAL PERFORMANCE CHARACTERISTICS

$V_{IN} = 5V$, $T_A = 25^\circ C$, unless otherwise noted.

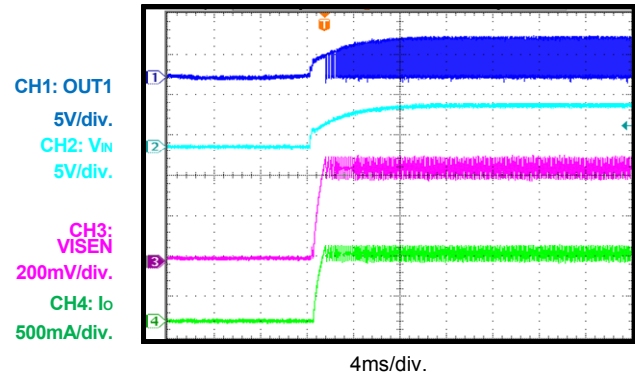
Normal Operation

IN1 = H, IN2 = L, $V_{IN} = 5V$, load = $2\Omega + 700\mu H$



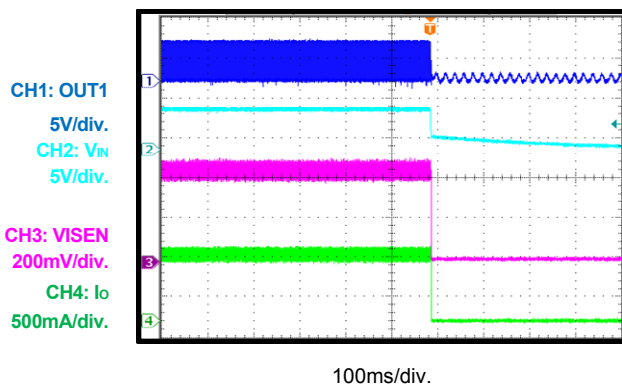
VIN Start-Up

IN1 = H, IN2 = L, $V_{IN} = 5V$, load = $2\Omega + 700\mu H$



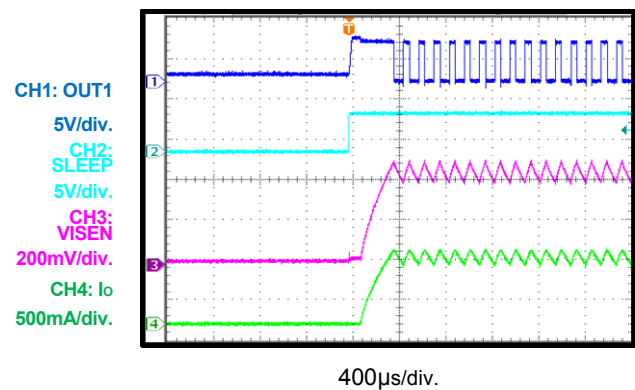
VIN Shutdown

IN1 = H, IN2 = L, $V_{IN} = 5V$, load = $2\Omega + 700\mu H$



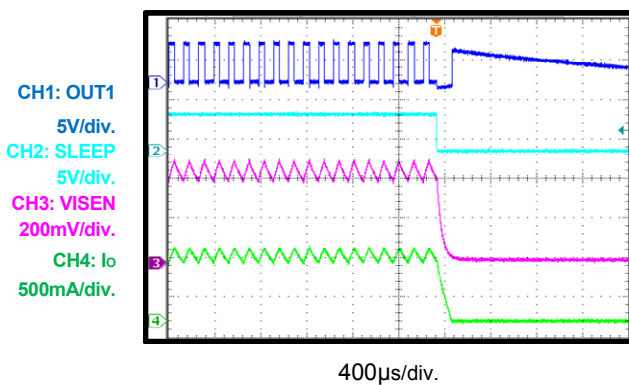
Sleep Start-Up

IN1 = H, IN2 = L, $V_{IN} = 5V$, load = $2\Omega + 700\mu H$



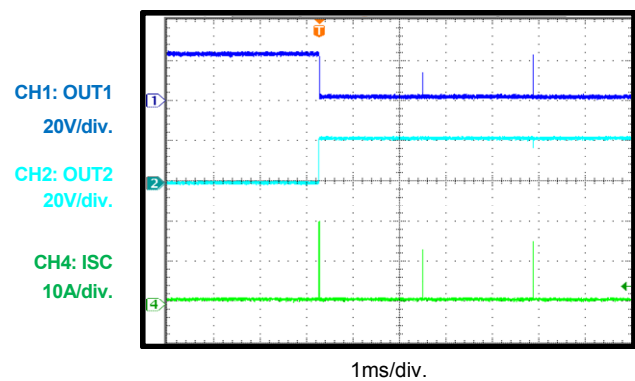
Sleep Shutdown

IN1 = H, IN2 = L, $V_{IN} = 5V$, load = $2\Omega + 700\mu H$



SCP-OUT Short to V_{IN}

IN1 = H, IN2 = L, $V_{IN} = 22V$

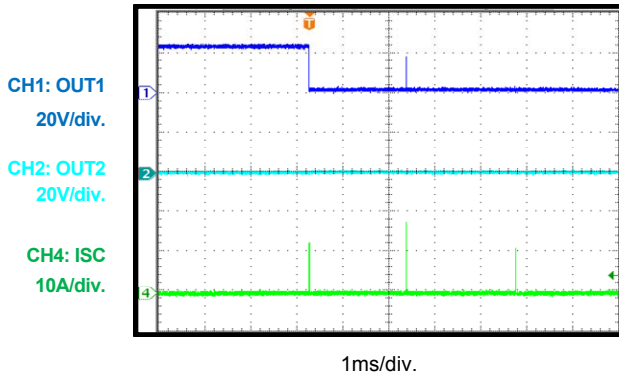


TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 5V$, $T_A = 25^{\circ}C$, unless otherwise noted.

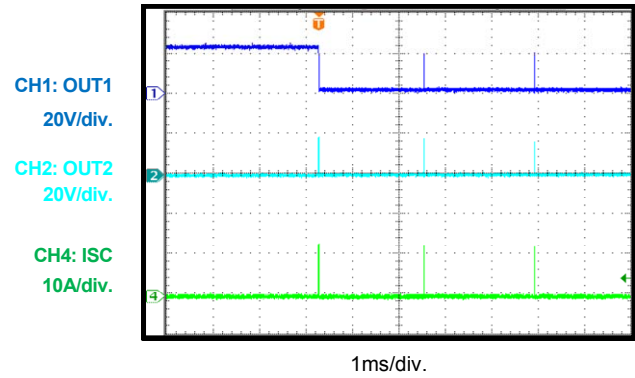
SCP-OUT Short to GND

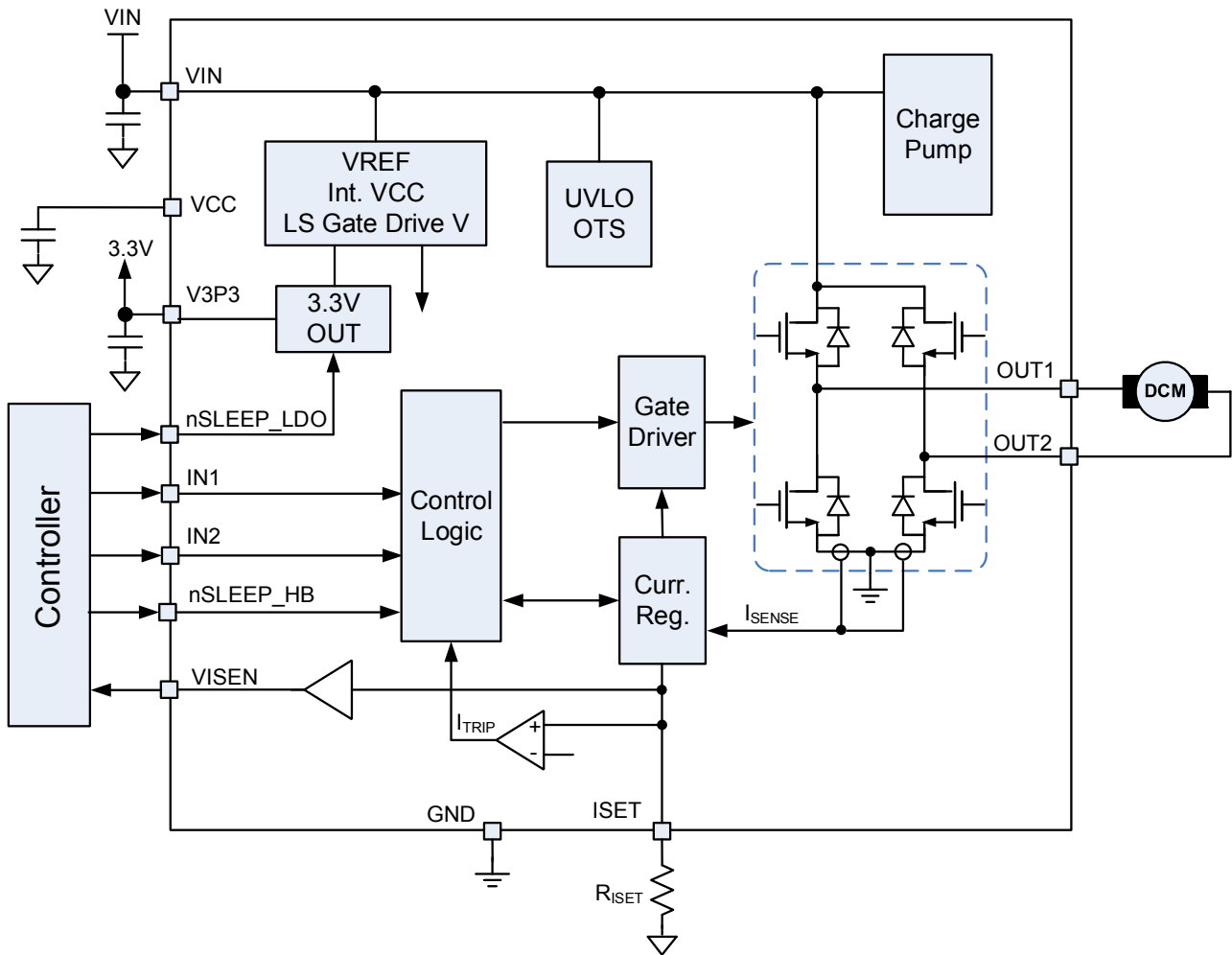
IN1 = H, IN2 = L, $V_{IN} = 22V$



SCP-OUT Short to OUT

IN1 = H, IN2 = L, $V_{IN} = 22V$



SINGLE-CHANNEL BLOCK DIAGRAM

Figure 2: Single-Channel Block Diagram

OPERATION

Bridge Control

The MP6550 is controlled using a PWM input interface, which is compatible with industry-standard devices. Each output is controlled by a corresponding input pin.

The following table shows the logic for the MP6550:

Table 1: Input Logic Truth Table

| IN1 | IN2 | OUT1 | OUT2 | Function (DC Motor) |
|-----|-----|------|------|---------------------|
| L | L | Z | Z | Coast |
| L | H | L | H | Reverse |
| H | L | H | L | Forward |
| H | H | L | L | Brake |

Table 2 shows the recommended PWM input control for a DC motor driver.

Table 2: PWM Input Control

| IN1 | IN2 | OUT1 | OUT2 | Function (DC Motor) |
|-----|-----|------|------|---------------------|
| H | PWM | L | H | Forward |
| | | H | L | Slow decay |
| PWM | L | H | L | Reverse |
| | H | | L | Slow decay |

Current Sensing

The current flowing in the two low-side MOSFETs is sensed with an internal current-sensing circuit. A voltage proportional to the output current is sourced on the VISEN pin.

The VISEN output voltage scaling is set by a resistor connected between the ISET pin and ground. For 1A of output current, 100µA of current is sourced into the resistor connected to VISEN. For example, if a 5kΩ resistor is connected between ISET and ground, the output voltage on the VISEN pin will be 0.5V/A of output current.

Current is sensed any time one of the low-side MOSFETs is turned on, including during slow decay (brake) mode.

The load current applied to the VISEN pin should be kept below 2mA, with no more than 500pF of capacitance.

Current Limit / Regulation

The current in the outputs is limited using constant-off-time PWM (pulse-width modulation) control circuitry. Figure 3 shows this operation.

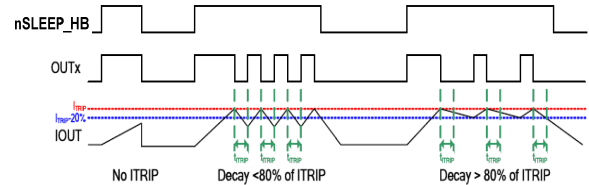


Figure 3: Current Limit Operation

Initially, a diagonal pair of MOSFETs turns on and drives current through the load. The current increases in the load, which is sensed by the internal current-sense circuit.

If the load current reaches the current trip threshold, the H-bridge switches to slow decay mode, and the two low-side MOSFETs turn on.

After a fixed off time (t_{ITRIP}), if the load current has fallen at least 20% below the current limit threshold, the FETs are re-enabled and the cycle repeats. If the current remains above this level, the off time is extended until the current falls to 20% below the current limit threshold.

The current limit threshold is reached when the VISEN pin reaches 0.5V. For example, with a 5kΩ resistor from ISET to ground, the VISEN voltage is about 0.5V/A of output current. So when the current reaches 1A, the VISEN pin voltage reaches 0.5V, and a current trip occurs.

During current regulation, the nFAULT pin is not activated.

Blanking Time

There is often a current spike during start-up due to the body diode's reverse-recovery current or the shunt capacitance of the load. This current spike requires filtering to prevent it from erroneously shutting down the high-side MOSFET. An internal fixed blanking time (t_{OCPD}) blanks the output of the current sense comparator when the outputs are switched. This blanking time also sets the minimum on time for high-side MOSFET.

nSLEEP Operation

The MP6550 has one H-bridge driver and a 3.3V reference output, which can be separately put into a low-power sleep state. Driving nSLEEP_HB low turns the H-bridge outputs off and disables all related internal circuits, including the gate drive charge pump. All inputs are ignored when nSLEEP_HB is active low. Driving nSLEEP_LDO low disables the 3.3V regulator. When waking up from sleep mode, some time needs to pass before the outputs will operate.

Protection Circuits

The MP6550 is fully protected against under-voltage, over-current, and over-temperature events.

Over-Current Protection (OCP)

The MP6550 has internal overload and short-circuit protection. The currents in both the high-

side and low-side MOSFETs are measured and if the current exceeds the current limit, all MOSFETs in the H-bridge are turned off. After approximately 1ms, the bridge is automatically re-enabled.

Thermal Shutdown (TSD)

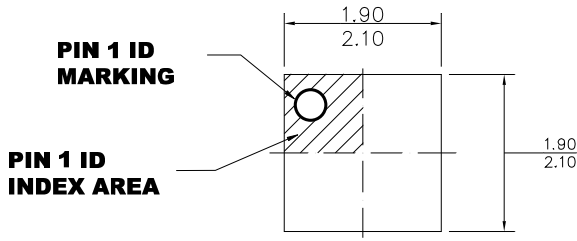
Thermal monitoring is also integrated into the MP6550. If the die temperature exceeds 160°C, all switches turn off. Once the die temperature has fallen to a safe level, operation resumes automatically.

Under-Voltage Lockout (UVLO)

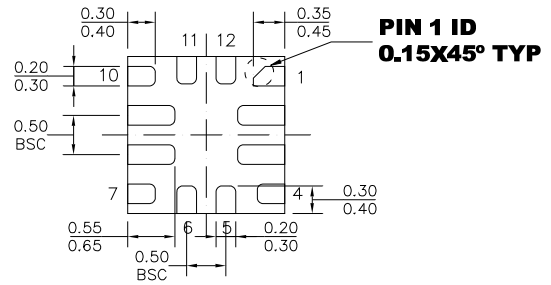
If at any time the voltage on the VIN pin falls below the under-voltage lockout (UVLO) threshold voltage, all circuitry in the device is disabled and the internal logic is reset. Normal operation resumes when VIN exceeds the UVLO threshold.

PACKAGE INFORMATION

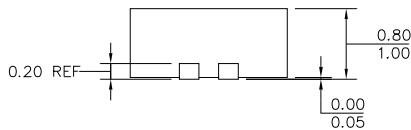
QFN-12 (2mmx2mm)



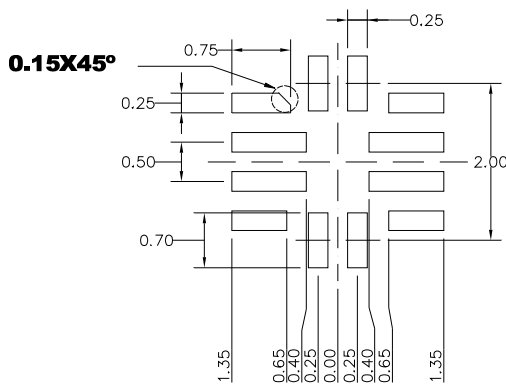
TOP VIEW



BOTTOM VIEW



SIDE VIEW



RECOMMENDED LAND PATTERN

NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) LEAD COPLANARITY SHALL BE 0.08 MILLIMETERS MAX.
- 3) JEDEC REFERENCE IS MO-220.
- 4) DRAWING IS NOT TO SCALE.

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