DESCRIPTION

The MP2625B is a monolithic switch mode battery charger with power path management for single-cell Li-ion batteries in a wide range of tablet and other portable devices. It integrates a synchronous BUCK regulator to provide regulated voltage for powering the system output and at the same time charging the battery. This device supports both USB and high power DC adapter input. In USB mode, the input current limit can be programmed to 450mA or 825mA via the logic pins to cover the USB2.0 and USB3.0 specifications. When the adapter input is present, the input current can also be limited in order to avoid overloading of the DC adapter. Input current limit can be programmed up to 2A.

The smart power path management allows MP2625B to regulate the system voltage for powering an external load and charging the battery independently and simultaneously. This allows immediate system operation even under missing or deeply discharged battery. When the input current limit is reached, the system load is satisfied in priority, then the charger will take the remaining current to charge the battery. Additionally, the smart power path control allows an internal connection from battery to the system in order to supplement additional power to the load in the event the system power demand increases over the input limited power or the input is removed.

The MP2625B features high integration with all the power switches integrated inside. No external MOSFET, blocking diodes, or current sense resistor is required.

Two status monitor output pins are provided to indicate the battery charge status and power source status. Other features include trickle charge, battery temperature monitoring, timer and thermal limiting regulation on chip.

MP2625B is available in QFN 3mmx4mm package.

FEATURES

- 4V to 10V Operating Input Voltage
- Smart Power Path Management
- Five Control Loops: Input Current Limit, Input Voltage Limit, Constant Charge Current, Terminal Battery Control and Thermal Fold-Back.
- 1.6MHz Switching Frequency
- Programmable Input Current Limit
- Programmable Charge Current
- Single Input for USB and AC adapter
- Cover USB2.0 and USB3.0 Input Specification
- Fully Integrated Power Switches
- No External Blocking Diode and Sense Resistor Required
- Charging Operation Indicator
- Built-in Programmable Charging Timer
- Thermal Limiting Regulation on Chip
- Battery Temperature Monitor
- Tiny Package Features Small Size

APPLICATIONS

- Smart Phone
- E-Book
- GPS
- Portable Media Player
- Portable Hand-held Solution
- Tablet PC

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TYPICAL APPLICATION

Efficiency vs. System Current
Charger disabled, V_SYS = 4.3V

<table>
<thead>
<tr>
<th>EFFICIENCY (%)</th>
<th>0</th>
<th>50</th>
<th>70</th>
<th>80</th>
<th>90</th>
<th>100</th>
</tr>
</thead>
<tbody>
<tr>
<td>I_SYS (A)</td>
<td>0</td>
<td>0.5</td>
<td>1</td>
<td>1.5</td>
<td>2</td>
<td></td>
</tr>
</tbody>
</table>

V_in = 6V
V_in = 3V
ORDERING INFORMATION

<table>
<thead>
<tr>
<th>Part Number*</th>
<th>Package</th>
<th>Top Marking</th>
</tr>
</thead>
<tbody>
<tr>
<td>MP2625BGL</td>
<td>QFN-20 (3mmx4mm)</td>
<td>See Below</td>
</tr>
</tbody>
</table>

* For Tape & Reel, add suffix –Z (e.g. MP2625BGL–Z);

TOP MARKING

`MPYN`

2625

BLLL

MP: MPS prefix
Y: year code
W: week code
2625B: part number
LLL: lot number

PACKAGE REFERENCE

TOP VIEW

QFN-20 (3mmx4mm)
### PIN FUNCTIONS

<table>
<thead>
<tr>
<th>Package Pin #</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>BST</td>
<td>Bootstrap. A capacitor is connected between SW and BST pin to form a floating supply across the power switch driver to drive the power switch’s gate above the supply voltage.</td>
</tr>
<tr>
<td>3, 4</td>
<td>IN</td>
<td>Power input of the IC from adapter or USB.</td>
</tr>
<tr>
<td>5</td>
<td>PGND</td>
<td>Power ground.</td>
</tr>
<tr>
<td>6</td>
<td>EN</td>
<td>Function logic control pin of the IC. Logic low to enable the part and logic high to disable the part.</td>
</tr>
<tr>
<td>7</td>
<td>M0</td>
<td>Mode Select Input Pin, in combination with M1 pin, setting the input current limit mode.</td>
</tr>
<tr>
<td>8</td>
<td>M1</td>
<td>Mode Select Input Pin, in combination with M0 pin, setting the input current limit mode.</td>
</tr>
<tr>
<td>9</td>
<td>CHGOK</td>
<td>Open drain output. It is pulled low during charging. And it is pulled high through an external resistor to VCC to indicate charge completed.</td>
</tr>
<tr>
<td>10</td>
<td>ACOK</td>
<td>Open drain output. It is pulled low to indicate the presence of a valid input power supply. Otherwise, it is pulled high through an external resistor to VCC to indicate invalid input or removed input.</td>
</tr>
<tr>
<td>11</td>
<td>AGND</td>
<td>Analog ground.</td>
</tr>
<tr>
<td>12</td>
<td>SYSFB</td>
<td>SYS voltage program pin. Connect a resistor divider from the pin to SYS and AGND to program the system output voltage. Leave the pin float to disable the function.</td>
</tr>
<tr>
<td>13</td>
<td>SYS</td>
<td>DC-DC regulator output to power the system load and charge the battery.</td>
</tr>
<tr>
<td>14</td>
<td>BATT</td>
<td>Positive battery terminal.</td>
</tr>
<tr>
<td>15</td>
<td>ISET</td>
<td>Charge current program pin. A resistor from the pin to AGND can program the charge current during CC charge. Float the pin will disable the charge function.</td>
</tr>
<tr>
<td>16</td>
<td>NTC</td>
<td>Thermistor input. Connect a resistor from this pin to VCC and the thermistor from this pin to ground. The thermistor is usually inside the battery pack.</td>
</tr>
<tr>
<td>17</td>
<td>ILIM</td>
<td>Input current limit program pin. A resistor from the pin to AGND can program the input current limit with adapter input.</td>
</tr>
<tr>
<td>18</td>
<td>TMR</td>
<td>Set timer out period. Connect TMR pin to AGND to disable the internal timer.</td>
</tr>
<tr>
<td>19</td>
<td>VLIM</td>
<td>Input voltage clamp program pin.</td>
</tr>
<tr>
<td>20</td>
<td>VCC</td>
<td>Supply voltage of the IC.</td>
</tr>
</tbody>
</table>
ABSOLUTE MAXIMUM RATINGS (1)
IN, SW .......................................... -0.3V to +20V
BATT, SYS ....................................... -0.3V to +6V
BST ............................................. -0.3V to +26V
All Other Pins ............................... -0.3V to +6V
Continuous Power Dissipation (T_A = +25°C) (2)
QFN20 3mmx4mm .............................. 2.6W
Junction Temperature ......................... 150°C
Lead Temperature ............................ 260°C
Storage Temperature ........................ -65°C to 150°C

Thermal Resistance (4) θ_JA θ_JC
QFN-20 (3mmx4mm) ....................... 48 ....... 11 °C/W

Notes:
1) Exceeding these ratings may damage the device.
2) The maximum allowable power dissipation is a function of the maximum junction temperature (T_J (MAX)), the junction-to-ambient thermal resistance θ_JA, and the ambient temperature (T_A). The maximum allowable continuous power dissipation at any ambient temperature is calculated by P_D (MAX) = (T_J (MAX)-T_A)/θ_JA. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
3) The device is not guaranteed to function outside of its operating conditions.
4) Measured on JESD51-7, 4-layer PCB.

Recommended Operating Conditions (3)
Supply Voltage V_IN ............................ 4.0V to 10V
Operating Junction Temp. (T_J) .......... -40°C to +125°C
## ELECTRICAL CHARACTERISTICS

$V_{IN} = 5.0V$, $T_A = 25^\circ C$, unless otherwise noted.

<table>
<thead>
<tr>
<th>Parameters (IN)</th>
<th>Symbol</th>
<th>Condition</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
</tr>
</thead>
<tbody>
<tr>
<td>Input Power (IN)</td>
<td>$V_{IN}$</td>
<td>Operating Range</td>
<td>4.0</td>
<td>10</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td>IN Under Voltage Lockout Threshold</td>
<td></td>
<td>Rising</td>
<td>3.65</td>
<td>3.8</td>
<td>3.95</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Falling</td>
<td>3.35</td>
<td>3.5</td>
<td>3.65</td>
<td>V</td>
</tr>
<tr>
<td>IN vs. BATT Threshold</td>
<td>$V_{BST-VSW}$</td>
<td>Rising</td>
<td>2.7</td>
<td>2.9</td>
<td>3.1</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Falling</td>
<td>2.55</td>
<td>2.75</td>
<td>2.95</td>
<td>V</td>
</tr>
<tr>
<td>BST Voltage Threshold</td>
<td></td>
<td>USB2.0 Mode</td>
<td>400</td>
<td>450</td>
<td>500</td>
<td>mA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>USB3.0 Mode</td>
<td>750</td>
<td>825</td>
<td>900</td>
<td>mA</td>
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<tr>
<td></td>
<td></td>
<td>Default Mode</td>
<td>1840</td>
<td>2000</td>
<td>2160</td>
<td>mA</td>
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<tr>
<td></td>
<td></td>
<td>Programmable Mode, $R_{ILIM}=23k$</td>
<td>1840</td>
<td>2000</td>
<td>2160</td>
<td>mA</td>
</tr>
<tr>
<td>Input Current Limit Reference Voltage</td>
<td>$V_{ILIM}$</td>
<td></td>
<td>1.1</td>
<td>1.14</td>
<td>1.18</td>
<td>V</td>
</tr>
<tr>
<td>High-side NMOS On Resistance</td>
<td>$R_{H_{DS(ON)}}$</td>
<td>Include the BLOCK FET</td>
<td>120</td>
<td>130</td>
<td>mΩ</td>
<td></td>
</tr>
<tr>
<td>Low-side NMOS On Resistance</td>
<td>$R_{L_{DS(ON)}}$</td>
<td></td>
<td>80</td>
<td>100</td>
<td>mΩ</td>
<td></td>
</tr>
<tr>
<td>High-side NMOS Peak Current limit</td>
<td></td>
<td></td>
<td>3.0</td>
<td>4.0</td>
<td>5.0</td>
<td>A</td>
</tr>
<tr>
<td>Input Voltage Clamp Threshold</td>
<td>$V_{VLIM}$</td>
<td>Voltage on VLIM</td>
<td>1.49</td>
<td>1.52</td>
<td>1.55</td>
<td>V</td>
</tr>
<tr>
<td>Input Quiescent Current</td>
<td>$I_{IN}$</td>
<td>Charger Enabled, USB2.0 Mode</td>
<td>2.4</td>
<td>5</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Charger Enabled, USB3.0 Mode</td>
<td>2.8</td>
<td>5</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Charger Enable, Programmable Mode</td>
<td>3.8</td>
<td>5</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Charger Enabled, Default Mode</td>
<td>3.8</td>
<td>5</td>
<td>mA</td>
<td></td>
</tr>
<tr>
<td></td>
<td></td>
<td>Disabled, EN=0V</td>
<td>3</td>
<td>5</td>
<td>uA</td>
<td></td>
</tr>
<tr>
<td>SYS to IN reverse current blocking</td>
<td></td>
<td>SYS=SW=4.5V, VIN=0V, monitor VIN leakage</td>
<td>0.01</td>
<td>0.2</td>
<td>uA</td>
<td></td>
</tr>
<tr>
<td>SYS Output</td>
<td>$V_{SYS}$</td>
<td>SYS voltage @ $V_{BATT}\leq 3.4V$, SYSFB float</td>
<td>3.45</td>
<td>3.6</td>
<td>3.75</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>$3.4V&lt;V_{BATT}\leq 4.2V$, SYSFB float, BATT Float</td>
<td>3.5</td>
<td>$V_{BATT}+ 0.2V$</td>
<td>4.5</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>User Programmed by SYSFB</td>
<td>4.08</td>
<td>4.4</td>
<td>V</td>
<td></td>
</tr>
<tr>
<td></td>
<td>$V_{SYS_REF}$</td>
<td></td>
<td>1.135</td>
<td>1.152</td>
<td>1.170</td>
<td>V</td>
</tr>
</tbody>
</table>
ELECTRICAL CHARACTERISTICS (continued)

$V_{IN} = 5.0V$, $T_A = 25^\circ C$, unless otherwise noted.

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Symbol</th>
<th>Condition</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
</tr>
</thead>
<tbody>
<tr>
<td>BATT to SYS Resistance</td>
<td></td>
<td>$V_{IN}=0V$, $I_{SYS}=200mA$, $V_{BATT}=4.2V$</td>
<td>0.04</td>
<td>0.05</td>
<td></td>
<td>Ω</td>
</tr>
<tr>
<td>BATT to SYS Current Limit</td>
<td></td>
<td>$V_{SYS}&gt;V_{BATT}-800mV$, $V_{BATT}=4.2V$</td>
<td>4</td>
<td>5</td>
<td>6</td>
<td>A</td>
</tr>
<tr>
<td></td>
<td></td>
<td>SYS short</td>
<td></td>
<td></td>
<td>230</td>
<td>mA</td>
</tr>
</tbody>
</table>

**Battery Charger**

| Terminal Battery Voltage       | $V_{BATT}$ | $V_{BATT}>V_{RECH}$, $I_{CHG}<|I_{BF}$, SYSFB float | 4.168 | 4.2  | 4.232 | V     |
|                                |           | $V_{SYS}<4.2V$ Programmed by SYSFB Pin           |      |      |      | V     |
| Recharge Threshold at $V_{BATT}$ | $V_{RECH}$ | SYSFB Float                                     | 3.9  | 4.0  | 4.1  | V     |
|                                |           | SYSFB programmed                                | 3.85 | 3.95 | 4.05 | V     |
| Recharge Hysteresis            |           |                                                  | 85   |      |      | mV    |
| Trickle Charge Threshold       |           |                                                  | 2.9  | 3    | 3.1  | V     |
| Trickle Charger Hysteresis     |           |                                                  | 200  |      |      | mV    |
| Trickle Charge Current         | $I_{TRICKLE}$ |                                              | 10%  |      |      | I_{CC} |
| Termination Charger Current    | $I_{BF}$  |                                                  | 5%   | 10%  | 15%  | I_{CC} |
| $I_{BF}$ Maximum Limit         |           |                                                  | 150  | 200  |      | mA    |
| Constant Current Mode Charge Current | $I_{CC}$ | $R_{ISET}=1.05k$                                 | 1.8  | 2.0  | 2.2  | A     |
|                                |           | $R_{ISET}=1.53k$                                 | 1.26 | 1.4  | 1.54 | A     |
|                                |           | $R_{ISET}=4.6k$                                  | 0.405 | 0.45 | 0.495 | A    |
| ISET Reference Voltage         |           |                                                  | 1.1  | 1.15 | 1.2  | V     |
| Battery UVLO                   |           | Rising                                          | 2.4  | 2.6  | 2.8  | V     |
|                                |           | Falling                                         | 2.2  | 2.4  | 2.6  | V     |
| Idea Diode Regulation Voltage  | $V_{SYS}$ | Supplement Mode                                 | $V_{BATT}-65mV$ |      |      | mV    |
| BATT Leakage Current           | $I_{BATT}$ |                                                  | 20   | 30   |      | μA    |

**ACOK, CHGOK**

| ACOK, CHGOK Pin Output Low Voltage | Sinking 5mA | 270 | 350 | mV |
| ACOK, CHGOK Pin Leakage Current   | Connected to 3.3V | 0.1 | 0.5 | μA |

**Timer**

| Trickle Charge Timer | $C_{TMR}=0.1μF$, $R_{ISET}=2.3k$ | 30  | 43  | 55 | Min |
| CC Charge Timer      | $C_{TMR}=0.1μF$, $R_{ISET}=2.3k$ | 5   | 6.2 | 7.4 | Hour |
### ELECTRICAL CHARACTERISTICS (continued)

*VIN = 5.0V, TA = 25°C, unless otherwise noted.*

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Symbol</th>
<th>Condition</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
</tr>
</thead>
<tbody>
<tr>
<td>Negative Temperature Coefficient (NTC) Control</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>NTC Low Temp Rising Threshold</td>
<td>VTHL</td>
<td>RNTC=NCP18XH103F 0°C</td>
<td>63</td>
<td>65</td>
<td>67</td>
<td>%VCC</td>
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<tr>
<td>Hysteresis on Low Temp Threshold</td>
<td></td>
<td></td>
<td>35</td>
<td></td>
<td></td>
<td>mV</td>
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<tr>
<td>NTC High Temp Falling Threshold</td>
<td>VTTH</td>
<td>RNTC=NCP18XH103F, 50°C</td>
<td>32</td>
<td>33.5</td>
<td>35</td>
<td>%VCC</td>
</tr>
<tr>
<td>Hysteresis on High Temp Threshold</td>
<td></td>
<td></td>
<td>70</td>
<td></td>
<td></td>
<td>mV</td>
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<tr>
<td>VCC Supply</td>
<td></td>
<td></td>
<td></td>
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<td></td>
<td></td>
</tr>
<tr>
<td>VCC UVLO</td>
<td></td>
<td>Rising</td>
<td>3.15</td>
<td>3.35</td>
<td>3.55</td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Falling</td>
<td>2.8</td>
<td>3</td>
<td>3.2</td>
<td>V</td>
</tr>
<tr>
<td>VCC Output Voltage</td>
<td></td>
<td>0mA&lt;VCC&lt;25mA, VIN=6V</td>
<td>4.3</td>
<td>4.5</td>
<td>4.6</td>
<td>V</td>
</tr>
<tr>
<td>VCC Output Current Limit</td>
<td></td>
<td></td>
<td>40</td>
<td></td>
<td></td>
<td>mA</td>
</tr>
<tr>
<td>Logic</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>EN Input Low Voltage</td>
<td></td>
<td></td>
<td>0.4</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>EN Input High Voltage</td>
<td></td>
<td></td>
<td>1.5</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>EN Input Current</td>
<td></td>
<td>EN=4V</td>
<td>4</td>
<td>8</td>
<td></td>
<td>μA</td>
</tr>
<tr>
<td></td>
<td></td>
<td>EN=0V</td>
<td>-0.5</td>
<td>-0.1</td>
<td></td>
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</tr>
<tr>
<td>M0, M1</td>
<td></td>
<td>Logic High</td>
<td>1.5</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Logic Low</td>
<td>0.4</td>
<td></td>
<td></td>
<td>V</td>
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<tr>
<td>Protection</td>
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<tr>
<td>Thermal Limit Temperature</td>
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<tr>
<td>Thermal Shutdown</td>
<td></td>
<td></td>
<td>150</td>
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<td>°C</td>
</tr>
</tbody>
</table>
TYPICAL PERFORMANCE CHARACTERISTICS

$V_{IN} = 5.0V$, $V_{BATT} =$ Full Range, Default Mode, $I_{IN Limit}=2A$, $V_{SYS}=4.4V$, $R6$ and $R7$ are float, $I_{CHG}=2A$, $V_{IN Clamp}=4.5V$, $L = 1.2 \, \mu H$, $T_A = +25^\circ C$, unless otherwise noted.
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 5.0\text{V}$, $V_{BATT} = \text{Full Range}$, Default Mode, $I_{IN} \text{ Limit}=2\text{A}$, $V_{SYS}=4.4\text{V}$, R6 and R7 are float, $I_{CHG}=2\text{A}$, $V_{IN} \text{ Clamp}=4.5\text{V}$, $L = 1.2 \mu\text{H}$, $T_A = +25^\circ\text{C}$, unless otherwise noted.
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 5.0V$, $V_{BATT} =$ Full Range, Default Mode, $I_{IN}$ Limit=2A, $V_{SYS}$=4.4V, R6 and R7 are float, $I_{CHG}$=2A, $L = 1.2 \, \mu H$, $T_A = +25^{\circ}C$, unless otherwise noted.

**CV Steady State**
- $V_{BATT} = 4.2V$, $I_{SYS} = 0A$

**CV Steady State**
- $V_{IN}$=9V, $V_{BATT}$=4.2V, $I_{SYS}$=1.5A

**Battery Full Steady State**
- $V_{BATT} = 4.2V$, $I_{SYS} = 1.5A$

**Battery Float Steady State**
- No Battery, $I_{SYS} = 1.5A$

**Input Current Limit**
- USB2.0 Mode, $V_{BATT}$ = 3.7V
  - Ramp up $I_{SYS}$

**Input Current Limit**
- $V_{IN}$=1A Adapter IN, $V_{BATT}$ = 4.2V
  - Ramp up $I_{SYS}$

**Input Voltage Limit**
- Battery Full, $I_{SYS}$ = 1-2A transient, 0.05A/µs

**System Current Transient**
- Battery Full, $I_{SYS}$ = 1-2A transient, 0.05A/µs
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 5.0\, V$, $V_{BATT} =$ Full Range, Default Mode, $I_{IN}$ Limit$=2A$, $V_{SYS}=4.4\, V$, $R6$ and $R7$ are float, $I_{CHG}=2A$, $L = 1.2\, \mu H$, $T_A = +25^{\circ}C$, unless otherwise noted.

**Plug in Adapter**

$V_{BATT} = 3.7\, V$, $I_{SYS} = 2A$

**Remove Adapter**

$V_{BATT} = 3.7\, V$, $I_{SYS} = 2A$

**Plug in USB**

USB 3.0 Mode, $V_{BATT} = 3.7\, V$

$I_{SYS} = 2A$

---

**Plug in Adapter**

No Battery, $V_{SYS} = 4.2\, V$, $I_{SYS} = 0.5A$

**Plug in Adapter**

$V_{IN} = 5V/1A$ Adapter,

$V_{BATT} = 3.7\, V$, $I_{SYS} = 0.5A$

**En On**

$V_{BATT} = 3.7\, V$, $I_{SYS} = 0A$

($V_{BATT} & V_{SYS}$ has 2.5V Offset)

---

**En Off**

$V_{BATT} = 3.7\, V$, $I_{SYS} = 0A$

($V_{BATT} & V_{SYS}$ has 2.5V Offset)

**En On**

$V_{BATT} = 3.7\, V$, $I_{SYS} = 1.5A$

($V_{BATT} & V_{SYS}$ has 2.5V Offset)

---

**Charge On**

$V_{IN} = 9V$, No Battery, $I_{SYS} = 0.5A$

---

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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN} = 5.0V$, $V_{BATT} =$ Full Range, Default Mode, $I_{IN}$ Limit=2A, $V_{SYS}=4.4V$, $R6$ and $R7$ are float, $I_{CHG}=2A$, $L = 1.2 \mu H$, $T_A = +25^\circ C$, unless otherwise noted.

- **Battery Insertion**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$

  ![Battery Insertion Graph](image)

- **Battery Insertion (USB 3.0 Mode)**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$

  ![Battery Insertion Graph](image)

- **Time Suspend @ NTC Fault**
  - $V_{BATT} = 3.8V$, $I_{SYS} = 0A$, $C_{TMR} = 165pF$

  ![Time Suspend Graph](image)

- **Timer Refreshed @ Battery Full**
  - $V_{BATT} = 4.2V$, $I_{SYS} = 0A$, $C_{TMR} = 165pF$

  ![Timer Refreshed Graph](image)

- **Thermal Shutdown**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$
  - Increase the Ambient Temperature slowly

  ![Thermal Shutdown Graph](image)

- **Indication @ Charge On/Off**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$

  ![Indication Graph](image)

- **Indication @ NTC Fault**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$

  ![Indication Graph](image)

- **Indication @ Timer Out**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$

  ![Indication Graph](image)

- **Indication @ Supplement**
  - $V_{BATT} = 3.7V$, $I_{SYS} = 0A$

  ![Indication Graph](image)
OPERATION

Figure 1: Function Block Diagram
Introduction

The MP2625B is a switching charger IC, with integrated smart power path management for powering the system and charging a single cell battery simultaneously and independently.

The MP2625B includes a high-voltage (up to 10V) input DC-DC step down converter for wide range of DC sources and USB inputs. It has precision average input current limit to make maximum use of the allowable input power. This feature allows fast charging when powering from an USB port, and ensures the input current never exceeds the input power specification especially when the input power comes from a USB port. Additionally, the input current limit threshold can be programmed by logic inputs or a resistor to ground from the ILIM pin.

The MP2625B implements an on-chip 40mΩ MOSFET which works as a full-featured linear charger with trickle charge, high accuracy constant current and constant voltage charge, charge termination, auto recharge, NTC monitor, built-in timer control, charge status indication, and thermal protection. The charge current can be programmed by an external resistor connected from the ISET pin to AGND. The IC limits the charge current when the die temperature exceeds 120°C.

The 40mΩ MOSFET works as an ideal diode to connecting the battery to the system load when the input power is not enough to power the system load. When the input is removed, the 40mΩ MOSFET is turned on allowing the battery to power up the system.

With smart power path management, the system load is satisfied in priority then the remaining current is used to charge the battery. The MP2625B will reduce charging current or even use power from the battery to satisfy the system load when its demand is over the input power capacity.

Figure 1 shows the function block diagram of MP2625B.

DC-DC Step Down Converter

The DC-DC converter is a 1.6MHz step-down switching regulator to provide the input power to the SYS, which drives the combination of the system load and battery charger.

The regulator contains input current measurement and control scheme to ensure the average input current remains below the level programmed via ILIM pin or logic inputs M0&M1. This meets the adapter capacity limit or stays in compliance with USB specification.

When the input voltage is higher than UVLO and 280mV higher than the battery voltage, input voltage OK signal is active (ACOK turns low) and the DC-DC converter soft-starts. If the input power is sufficient to supply the combination of the system load and battery charger, and the input current limit loop is not triggered. The converter output voltage $V_{SYS}$ will be regulated:

1) If $BATT>3.4V$, $V_{SYS}$ is approximately 0.2V above the battery voltage to minimize the power loss of the battery charger during fast charging.

2) If $BATT<3.4V$, $V_{SYS}$ is fixed at 3.6V to power the system immediately even when a drained battery is inserted to be charged. Figure 2 shows the relationship of $V_{SYS}$ vs. $V_{BATT}$.

System voltage can also be regulated to any value between 4.08V to 4.4V by using a resistor divider on the SYSFB pin. This is shown as R6 and R7 in Figure 10. If the SYSFB is left floating, the system program is invalid, and $V_{SYS}$ is regulated as Figure 2.

The converter adopts fixed off-time control to extend the duty cycle (close to 100%) when the input of the converter is close to $V_{SYS}$.
Close to 100% duty operation, BST refresh operation makes sure the driver voltage of the HS will be charged by turning on the LS until negative IL hit to a threshold. If the input power is insufficient to supply the combination of the system load and battery charger, the DC-DC converter will limit the total power requirement by restricting the input voltage, input current and the peak current through the MOSFET. The power path management will reduce the charge current to satisfy the external system load in priority. According to this feature, the USB specification is always satisfied first. Even if the charge current is set larger than the USB input current limit, the real charge current will be reduced as needed.

**Input Limit State**

If the input power is insufficient to supply the combination of the system load and battery charger, the MP2625B implements three input limit control loops to reduce the charge current and satisfy the external system load in priority. The input in this case might be limited as follows: input current limit, input voltage limit and DC-DC peak current limit.

**Input Current Limit:** When the input current is higher than the programmed input current limit the input current limit loop takes the control of the converter and regulates the input current at constant value. When the battery voltage is over 3.4V, the output voltage (V_{SYS}) will drop down according to the increase of the system current, and the charge current drops down after the BATT-to-SYS switch (40mΩ MOSFET) is fully on according to V_{SYS} dropping down. During this process, the system voltage is slightly higher than V_{BATT}. When the battery voltage is lower than 3.4V, to maintain the minimum system voltage and ensure the system operation, the input current limit control will pull down the charge current directly to reduce the load of the converter so that the system current is satisfied in priority.

**Input Voltage Limit:** A resistor divider from IN pin to VLIM pin to AGND is used for the input voltage limit control. When the voltage on VLIM pin hits the reference voltage of 1.52V, the output of the input voltage limit error amplifier will drop in to control the operation duty. In this mode, the input voltage will be clamped according to the value set by the resistor divider. The control to the system voltage and charge current is the same as the one explained in the input current limit. Charge current drops down to satisfy the system current request first. This feature provides a second protection to the input power and ensures the safe operation of the input adapter. Even if a wrong adapter is inserted, the MP2625B can continue operation, providing the maximum power to its load. User can program the input voltage limit value through the resistor divider from IN to VLIM to AGND.

**Peak Current Limit:** The peak current of the high side switch of the DC-DC converter is sensed during every cycle, it is compared to the reference 4A, if the peak current hits the threshold, the peak current limit mode is triggered. The control of the charge current is the same with the above two limits.

**Input Current Limit Setting**

The current at ILIM is a precise fraction of the adapter input current. When a programming resistor is connected from ILIM to AGND, the voltage on ILIM represents the average input current of the PWM converter. And the input current approaches the programmed limit, ILIM voltage reaches 1.14V.

The average input current limit can be set through the resistor connecting from ILIM to AGND according to the following expression:

\[
I_{IN\_LIMIT} = 1.14 \times \frac{40000}{R_{ILIM}\ (k\Omega)} (mA)
\]

When USB input, the input current limit is set internally and the programmed value is invalid. The MP2625B provides typical of 450mA input current limit for USB2.0 specification and a typical of 825mA for USB3.0 specification respectively.

The user can choose to set the input current limit through the two logic pins M0 and M1 as shown in Table 2 according to its input specification.

When both M0 and M1 are float, they are pulled to the logic high, under this condition, the input current is limited to a default value of 2A.
Table 1: Input Current Limit Setting

<table>
<thead>
<tr>
<th>M0</th>
<th>M1</th>
<th>Mode</th>
</tr>
</thead>
<tbody>
<tr>
<td>Low</td>
<td>Low</td>
<td>USB2.0 Mode</td>
</tr>
<tr>
<td>Low</td>
<td>High</td>
<td>USB3.0 Mode</td>
</tr>
<tr>
<td>High</td>
<td>Low</td>
<td>Programmable Mode</td>
</tr>
<tr>
<td>High/Float</td>
<td>High/Float</td>
<td>Default Mode</td>
</tr>
</tbody>
</table>

Input Voltage Limit Setting

The input voltage can be limited at a value set by a resistor divider from IN pin to VLIM pin to AGND according to the following expression (Typical Application Circuit):

\[
V_{IN,LIM} = 1.52 \times \frac{R1 + R2}{R2} (V)
\]

When the voltage on VLIM pin drops and hits the reference voltage 1.52V, the input voltage will be clamped to the setting value.

It is recommended to set the VIN_LIM not below the input voltage UVLO rising threshold, and 320mV higher than the battery voltage.

Battery Charger

The MP2625B completes charge operation consist of trickle charge, automatic charge termination, charge status indication, timer control, NTC indication, automatic recharge, and thermal limiting.

When the PWM converter is out of soft start, the battery charge cycle begins, the MP2625B first determines if the battery is deeply discharged. If the battery voltage is lower than the trick charge threshold (typical 3.0V), the battery charger starts in “trickle charge mode”. The trickle charge current is limited to 10% of the programmed charge current until the battery voltage reaches 3.0V. If the charge stays in the “trick charge mode” for longer than “trick charge timer period”, the “timer out” condition is triggered, the charge is terminated and CHGOK will start blinking to indicate the fault condition.

If the charge current drops below I_{BF} because of any limit condition, the MP2625B will come out of CV mode, and the charge full detection is invalid.

A safe timer starts at the beginning of each new charge cycle and it monitors if the whole charge period is within the programmed time limit. After each charge cycle, when the battery is indicated as full, the timer counter will be reset. If the time is expired while the charging is still on going, the timer will force the MP2625B to terminate charging CHGOK is blinking to indicate the fault condition.

If system voltage is programmed lower than 4.2V by the resistor divider at the SYSFB pin, the battery will be charged most close to \(V_{SYS}\) until the charge current reaches the \(I_{BF}\) threshold.

Automatic Recharge

Once the battery charge cycle is completed, the MP2625B turns off indicating the battery full status. During this process, the battery power may be consumed by the system load or self-discharge. If the input power is always on, to ensure the battery not to be exhausted, the new charge cycle will automatically begin when the battery voltage falls below the auto-recharge threshold \(V_{RCCHG}\) which is typically 4V when the SYSFB is float, and 50mV lower if the SYSFB is connected to a resistor divider. The timer will restart when the auto-recharge cycle begins.

During the charge off state when the battery is fully charged, if the input power is recycled, or
the EN signal is refreshed, the charge cycle will re-start and the timer will refresh even if the battery voltage is above the auto-recharge threshold.

Charge Current Setting
The charge current of the MP2625B is programmed using a single resistor from ISET pin to ground. The program resistor and charge current are calculated using the following equations:

\[ I_{CHG} = 1.15 \times \frac{1800}{R_{SET} \Omega} \text{(mA)} \]

At either constant current mode or constant voltage mode, the voltage at the ISET pin is proportional to the actual charge current delivered to the battery, IBATT. The battery current can be calculated by monitoring the ISET pin voltage with the following formula:

\[ I_{BATT} = \frac{V_{ISET}}{1.15} \times I_{CHG} \]

Additionally, the actual battery charge current may be lower than the programmed current due to limited input power available and prioritization of the system load.

Battery charge full current threshold \( I_{BF} \) is set internally at 10% of the programmed charge current. However, \( I_{BF} \) has a 150mA maximum limit which will not be exceeded.

Ideal Diode Mode
If the system current requirement increases over the preset limit of the PWM converter, the additional current will be drawn from the battery via the BATT-to-SYS switch. To avoid very large currents being drawn from the battery which might affect the reliability of the device, the MP2625B controls the charge switch to work at the ideal diode mode regulating \( V_{SYS} \) to \( V_{BATT} - 65\text{mV} \) when \( V_{SYS} \) is 40mV lower than \( V_{BATT} \) is detected. Only when \( V_{SYS} \) is 40mV higher than \( V_{BATT} \), the charger switch exits the ideal diode mode, and the charge cycle softly restarts.

Logic Control
The MP2625B has two separate enable control pins.

EN is a logic control pin that controls the operation of the whole IC. When EN is low, the IC is enabled and the PWM converter output powers the system and the charger. When EN is high, both the PWM converter and the charger are disabled. The BATT to SYS switch turns fully on to connect the battery to power the system.

The ISET pin can be also used to control the operation of the charger. Setting ISET pin floating will disable the charger function while the output of PWM converter will continue supply power to system. On the other hand, a resistor from ISET to AGND will enable the charging at the programmed charge current.

The logic control of the ISET pin of the MP2625B can be realized as Figure 3. In this way, the user can choose logic low to be “off” signal or logic high to be “on” signal with a N-MOSFET.
Input Power Status Indication (ACOK)
An internal under voltage lockout circuit monitors the input voltage and keeps the IC in off state until the input rises over the rising threshold (3.8V). When the input voltage decreases below threshold (3.5V), the IC will turn off, and the system load will be powered by the battery automatically. ACOK is an open-drain, active-low output that indicates the status of input power. The input is considered valid when the input voltage is over the UVLO rising threshold, and 280mV higher than the battery voltage to ensure both the converter and the charger can operate normally. If the input voltage from an adapter or from a USB port is indicated OK, ACOK will turn low.

During EN off or thermal shutdown conditions, the ACOK turns high to indicate no power is provided by the input to the system. The ACOK signal indicates if input supplies power to the system load or not. Any other condition can not affect the ACOK indication as long as the input power is present.

Charge Status Indication (CHGOK)
CHGOK is an open-drain, active-low output that indicates the status of charge. CHGOK will be low during normal charging operation, turn high after charge full, and blink if a fault condition happens including NTC fault (battery temperature invalid) and timer out (bad battery).

In the event of a fault condition, CHGOK switches at 6Hz with the 50% duty cycle and enter “blinking” mode. The user should check the application circuit to find out the root cause of the fault condition if the “blinking” signal is asserted.

For no battery condition, CHGOK is blinking according to the transition between charging and charge full. The blinking frequency is determined by the cycle of charge and discharge of the output capacitor.

When the charge current to the battery is low or in the event the battery is in supplement mode caused by the insufficient input power, CHGOK keeps low to avoid providing false charge full indication. Table 2 shows the ACOK and CHGOK status under different charge conditions.

<table>
<thead>
<tr>
<th>ACOK</th>
<th>CHGOK</th>
<th>Charger Status</th>
</tr>
</thead>
<tbody>
<tr>
<td>low</td>
<td>low</td>
<td>In charging, supplement mode</td>
</tr>
<tr>
<td>low</td>
<td>high</td>
<td>End of charge, ISET disable charger only.</td>
</tr>
<tr>
<td>low</td>
<td>blinking at 6Hz</td>
<td>NTC fault, timer out</td>
</tr>
<tr>
<td>high</td>
<td>high</td>
<td>V_{IN} absent, EN disable, thermal shutdown</td>
</tr>
</tbody>
</table>

Timer Setting
The MP2625B uses an internal timer to terminate charge if the timer times out. The timer duration is programmed by an external capacitor at the TMR pin and related to the real battery current (I_{BATT}).

\[ t_{\text{Cycle}} = \frac{C_{\text{TMR}} \times 0.73 \times 2^{\frac{I_{\text{BATT}}}{40000}} + 1 \times 10^{-6}}{s} \]

The trickle mode charge time is:

\[ t_{\text{Trickle,TMR}} = 45055 \times t_{\text{Cycle}} (s) \]

If a C_{TMR} (0.1\mu F) is connected, the trickle charge current is just 0.1A, and suppose the capacitance and battery current is fixed at this value, the trickle charge time is calculated about 31 minutes.

The total charge time (during CC and CV mode) is:

\[ t_{\text{Total,TMR}} = 3407871 \times t_{\text{Cycle}} (s) \]

If a C_{TMR} (0.1\mu F) is connected, the CC charge current is 1A, and suppose the capacitance and battery current is fixed at this value, the CC charge time is calculated about 5.3 hours.

During CV mode, as the charge current decrease naturally, so the timer cycle increases naturally according to the formula. The time period also will increased during CV mode. The battery is closer to full state, the time period is longer.

As a result of power path management control, charge current might vary during normal operation, under this condition, the MP2625B automatically takes into account this variation and adjust the timer period accordingly.
If charge is stopped due to high system load, the timer is temporarily suspended.

This feature avoids indicating a false trigger indication for bad battery indication when there is little charge current delivered to the battery as a result of the insufficient input power.

When the timer out condition occurs, the MP2625B terminates the charge at once and CHGOK blinks to indicate the fault status. If one of the following events happens, the timer is refreshed and MP2625B re-starts the charge cycle.

- Input re-startup
- Refresh EN /ISET signal
- Auto-Recharge

**Battery Temperature Monitor**

The MP2625B continuously monitors battery’s temperature by measuring the voltage at the NTC pin, typically determined by a negative temperature coefficient (NTC) thermistor and external voltage dividers, as shown in Figure 5. Both the external voltage divider and the internal reference voltage divider are pulled up to the same power pin. So actually, the voltage ratios between the internal and external dividers are compared. For the NTC thermistor, the hotter ambient temperature, the lower resistance is, which corresponds to a lower voltage ratio, vice versa.

\[
V_{\text{THL}} = \frac{V_{\text{TH,COLD}}}{V_{\text{CC}}} R_{\text{NTC,LOW}} = \frac{R_{T2}}{R_{T1} + R_{T2}}
\]

\[
V_{\text{THH}} = \frac{V_{\text{TH,HOT}}}{V_{\text{CC}}} R_{\text{NTC,HIGH}} = \frac{R_{T2}}{R_{T1} + R_{T2}}
\]

V_{\text{THL}} actually means the low temperature threshold limit ratio (Cold Limit) and V_{\text{THH}} means the high temperature threshold limit ratio (Hot Limit). R_{\text{NTC,LOW}} means the resistance of the selected NTC thermistor under low temperature, while R_{\text{NTC,HIGH}} means the resistance of the selected NTC thermistor under high temperature.

Once the external voltage ratio exits the low temperature threshold \( (V_{\text{THL}}) \) to high temperature threshold \( (V_{\text{THH}}) \) range, it means the battery temperature is out of the COLD to HOT range, the IC suspends charging and reports the NTC fault. Charging resumes automatically after the battery temperature is within the COLD to HOT temperature range again. Figure 6 shows the standard control.

**Thermal Protection**

The MP2625B implements thermal protection to prevent the thermal damage to the IC or surrounding components. An internal thermal sense and feedback loop will automatically decrease the charge current when the die-temperature rises to about 120°C. This function is referred as charge current thermal fold-back. This feature protects the MP2625B from excessive temperature due to high power operation or high ambient thermal conditions. Another benefit of this feature is charge current can be set according to the requirement rather
than worst-case conditions for a given application with the assurance of safe operation. The MP2625B will stop charging if the junction temperature rises above 150°C as the IC enters thermal shutdown protection.

**Battery Discharge Protection**

When the input power is removed or invalid, the system load will draw power from the battery via the battery switch. Under this condition, the battery switch is fully on to minimize the power loss. The MP2625B integrates battery discharge protection. If the battery discharge current is larger than the discharge current limit threshold $I_{DIS}$ (5A), the current will be regulated at the preset limited value. And if the current increases further, the SYS voltage starts to decrease. When $V_{SYS}$ drops to about 800mV lower than $V_{BATT}$, SYS short condition is detected. Under this condition, the discharge current is limited at 230mA. In the event of a short from system to GND, the discharge current from the battery to the system is also limited to 230mA.

Furthermore, battery voltage UVLO is always monitored. If the battery voltage is lower than the battery UVLO threshold, the battery switch is turned off immediately. This feature makes sure the battery from over-discharged.

**Dynamic Power Path Management (DPPM)**

In the presence of a valid input source, the PWM converter will supply the current to both the system and the battery charger. The voltage $V_{SYS}$ is regulated based on the value of the battery voltage. When $V_{BATT}$ is higher than 3.4V, $V_{SYS}$ is regulated 200mV above $V_{BATT}$ to charge the battery. When $V_{BATT}$ is lower than 3.4V, to ensure the system can still be powered up even with a drained battery connected, $V_{SYS}$ is regulated at constant 3.6V.

When the input source is overloaded, either the current exceeds the input current limit or the voltage falls below the input voltage limit, the MP2625B then reduces the charge current until the input current falls below the input current limit and the input voltage rises above the input voltage limit. If the system current increases beyond the power allowed by the input source, additional power will be drawn from the battery via an on-chip 40mΩ MOSFET working as an ideal diode.

For example, in an application, when the input current is selected as USB3.0 Mode, (M0=Low, M1=High) which means the input current limit is set as 0.9A. When we adding the system load current, when the input current limit is hit, the charge current drops down to satisfy the system current. Even, when the system current increased further when the charge current drops to 0A, the battery will discharge to supplement the system load. The test waveform “Input Current Limit” is shown in the “TYPICAL PERFORMANCE CHARACTERISTICS”.

From the waveform, the input current is limited constantly. By this way, when set the input current limit below the input source capacity in the application, the input voltage will always be stable.

If a power source which has low capacity than the input current limit, then MP2625B uses VIN limit loop to protect the input voltage from crashing.

For example, when input current limit is set as 2A, while a USB3.0 power source is insert, then if system load is rising up, and the input current quickly overs 900mA, the input voltage drops down quickly while the input voltage limit loop limit it to a setting clamp value. And charge current drops down accordingly. Also, supplement mode is valid too.

The test waveform “Input Voltage Limit” is shown in the “TYPICAL PERFORMANCE CHARACTERISTICS”.

Additionally, if the input source is removed, the MP2625B will turn on the 40mΩ MOSFET allowing the battery to power the system load to keep the operation of the portable device.

**Operation Flow Chart**

Figure 7 shows the operation flow chart of the MP2625B.

Figure 8 shows the operation process.
Any Limit condition triggered?

"VBATT>3.0V?"

CC/CV Charge

"ICHG=IBF?"

Charge Full, EOC=1

TMR off, clear the counter

DC-DC keeps work

"VBATT<VRCHG?"

"VSYS<VBATT-40mV?"

"ICHG=0?"

Trickle Charge

ICHG=10%ICC

"Clamp DC-DC EAO to regulate the part at the limit state"

"V_{SYS}<V_{BATT}-40mV?"

"V_{BATT}<V_{OC}OL?"

"V_{SYS}>V_{BATT}+40mV?"

"V_{SYS}>V_{BATT}+65mV?"

"V_{BATT}>V_{BATT}_{UVLO}?"

"V_{SYS}>V_{BATT}+40mV?"

"V_{BATT}<V_{OC}OL?"

"Satisfy System current"

"Decrease ICHG, Keep V_{SYS}=3.6V"

"Disable Ideal Diode Mode"

"Ideal Diode Mode"

Battery switch shuts down, DC-DC in over load condition, V_{SYS} drops down

Figure 7: MP2625B Operation Flow Chart under No Fault Condition
Figure 8: MP2625B Operation Process under No Fault Condition
APPLICATION INFORMATION
COMPONENT SELECTION

Setting the Input Current Limit

First the input current limit can be set by the M0 and M1 pins refer to the Table 1, the exact current value in minimum, typical and maximum is listed in the EC table.

Under program mode, connect a resistor from the ILIM pin to AGND to program the input current limit for different input ports. The relationship between the input current limit and setting resistor is as Equation (1) which is shown in following again:

\[ I_{\text{IN,LIM}} = 1.14 \times \frac{40000}{R_{\text{ILIM}}} \text{(mA)} \]  

(1)

For MP2625B, the \( R_{\text{ILIM}} \) is greater than 22.8k, so that \( I_{\text{ILIM}} \) is not over 2A. The tolerance is ±8% of the input current limit setting.

So for a required minimum input current limit value, just calculate its typical value first, then calculate the setting resistor based on Equation (1). Also the maximum value can be calculated according to the tolerance. 1% accuracy resistor is used for this setting. Also, for a given resistor of \( R_{\text{ILIM}} \), the input current limit can be calculated. Following table is an example:

<table>
<thead>
<tr>
<th>Resistor</th>
<th>( R_{\text{ILIM}} ) (kΩ)</th>
<th>( I_{\text{IN,LIM}} ) (mA)</th>
<th>8%</th>
<th>-8%</th>
</tr>
</thead>
<tbody>
<tr>
<td>Typ.</td>
<td>54.9</td>
<td>830.601</td>
<td>897.049</td>
<td>764.153</td>
</tr>
<tr>
<td>Min.</td>
<td>54.351</td>
<td>838.991</td>
<td>906.11</td>
<td>771.872</td>
</tr>
<tr>
<td>Max.</td>
<td>55.449</td>
<td>822.377</td>
<td>888.168</td>
<td>756.587</td>
</tr>
</tbody>
</table>

Therefore, if customer selected a 54.9k in 1% accuracy resistor for the input current limit setting, then the typical input current limit value is 830.6mA, the minimum is 756.6mA and the maximum is 906mA.

Setting the Charge Current

\( R_{\text{SET}} \) connecting from the ISET pin to AGND sets the charge current (I\text{CHG}). The relationship between the charge current and setting resistor is as Equation (2) which is shown in following again:

\[ I_{\text{CHG}} = 1.15 \times \frac{1800}{R_{\text{SET}}} \text{(mA)} \]  

(2)

For example, if the typical I\text{CHG} is designed as 2A, then the \( R_{\text{SET}} \) is calculated at 1.05kΩ. The tolerance of the I\text{CHG} setting is ±10%. If the minimum or maximum charge current is required, first the typical value should be calculated according to the tolerance. After that, calculate the resistor according to formula (2). 1% accuracy resistor is used for this setting.

For a given setting resistor, the charge current can be calculated by the same way did in the input current limit setting. Usually in USB mode, the charge current is always set over the USB input limit specification. Then the MP2625B regulates the input current constant at the limitation value. Thus the real CC charge current is not the setting value, it varies with different input and battery voltages.

The maximum CC charge value can be calculated as:

\[ I_{\text{CC,MAX}} = \frac{V_{\text{IN}} \times I_{\text{ILIM}} \times \eta}{V_{\text{TC}}} \text{(A)} \]  

(3)

Where \( V_{\text{TC}} \) is trickle charge threshold (3V) and \( \eta \) is the current charge efficiency. Assume \( V_{\text{IN}}=5.5\text{V}, \ I_{\text{ILIM}}=1\text{A}, \) suppose \( \eta=83\% , \) thus \( I_{\text{CC,MAX}}=1.52\text{A}. \)

Figure 9 shows a calculating charge current curve by limiting the input current limit.
\[ V_{\text{VLIM}} = V_{\text{IN,\text{Lim}}} \times \frac{R_2}{R_1 + R_2} \] (4)

Thus the input voltage is:
\[ V_{\text{IN,\text{Lim}}} = \frac{R_1 + R_2}{R_2} V_{\text{VLIM}} \] (5)

The voltage clamp reference voltage \( V_{\text{VLIM}} \) is 1.52V, and a typical value for \( R_2 \) can be 10kΩ. With this value, \( R_1 \) can be determined by:
\[ R_1 = R_2 \times \frac{V_{\text{VLIM}} - V_{\text{IN,\text{Lim}}}}{V_{\text{VLIM}}} \] (6)

For example, for a 4.65V input limit voltage, \( R_2 \) is 10kΩ, and \( R_1 \) is 20.6kΩ.

The minimum value and the maximum value of the input voltage limit can be calculated according to the accuracy of the resistor and the tolerance of \( V_{\text{VLIM}} \). 1% accuracy resistors are used for \( R_1 \) and \( R_2 \).

**Setting the System Voltage**

The system voltage can be regulated to any value between 4.08V to 4.4V by the resistor divider on SYSFB pin as \( R_6 \) and \( R_7 \) in Figure 11.

\[ V_{\text{SYS}} = V_{\text{SYS,\text{Ref}}} \times \frac{R_6 + R_7}{R_7} \] (7)

Where \( V_{\text{SYS,\text{Ref}}} \) is 1.152V, the reference voltage of SYS. With a typical value for \( R_7 \), 10kΩ, \( R_6 \) can be determined by:
\[ R_6 = R_7 \times \frac{V_{\text{SYS}} - V_{\text{SYS,\text{Ref}}}}{V_{\text{SYS,\text{Ref}}} - V_{\text{SYS}}} \] (8)

For example, for a 4.2V system voltage, \( R_7 \) is 10kΩ, and \( R_6 \) is 26.5kΩ. 1% resistors are selected for the \( R_5 \) and \( R_6 \).

Be noted that, the minimum \( V_{\text{SYS}} \) is limited to be higher than the maximum value of the auto-recharge threshold which is 4.05V.

**Selecting the Inductor**

Inductor selection trades off among cost, size, and efficiency. A lower inductance value corresponds to a smaller size, but results in higher ripple currents, higher magnetic hysteretic losses, and higher output capacitances. From a practical standpoint, the inductor ripple current does not exceed 30% of the maximum load current under worst cases conditions. For example, if the \( I_{\text{CHG}} \) is setting to 2A in MP2625B, then, \( \Delta I \) is general set at 0.6A.

However, for the light load condition, the inductor ripple current will be very small which may cause unstable operation due to the peak current mode control of the IC. For stable operation, the experienced minimum limit value for inductor current ripple is 0.5A. Therefore, the inductor current ripple is the maximum one of 30% times \( I_{\text{CHG}} \) and 0.5A.

And the inductance can be calculated according to Equation (9):
\[ L = \frac{V_{\text{IN}} - V_{\text{SYS}}}{\Delta I_{\text{MAX}}} \times \frac{V_{\text{SYS}}}{\Delta I_{\text{MAX}}} (\mu\text{H}) \] (9)

The peak current of the inductor is calculated as Equation (10):
\[ I_{\text{PEAK}} = I_{\text{LOAD,\text{MAX}}} \times (1 + \frac{\% \text{ripple}}{2}) (\text{mA}) \] (10)

Where \( V_{\text{IN}} \), \( V_{\text{SYS}} \), and \( f_s \) are the typical input voltage, the output voltage, and the switching frequency, respectively.

Following Table 4 provides the selection guide of the inductance based on different input voltage.

<table>
<thead>
<tr>
<th>SPEC</th>
<th>Inductance Selection</th>
</tr>
</thead>
<tbody>
<tr>
<td>( V_{\text{IN}} )</td>
<td>( L = \frac{V_{\text{IN}} - V_{\text{SYS}}}{\Delta I} \times \frac{V_{\text{SYS}}}{V_{\text{IN}} \times f_s (\text{MHz})} ) (\muH)</td>
</tr>
<tr>
<td>5V</td>
<td>0.55</td>
</tr>
<tr>
<td>9V</td>
<td>2.25</td>
</tr>
</tbody>
</table>

**Table 4: Inductance Selection Guide under different Input Voltage**

**NOTE:**

5) Choose the inductor with a value a little lower than the calculated \( L_{\text{MIN}} \), makes the \( \Delta I \) increased a little, but the 2.2uH is more regular in the application which will have a lower cost.

6) Saturation Current of the inductor should be higher than the IPEAK, add 0.5A margin here.
Selecting the Input Capacitor

The input capacitor C1 from the typical application circuit absorbs the maximum ripple current from the PWM converter, which is given by

\[
I_{\text{RMS_MAX}} = I_{\text{CC_MAX}} \times \sqrt{V_{\text{TC}} \times (V_{\text{IN_MAX}} - V_{\text{TC}})} \quad (A) \quad (11)
\]

For \( I_{\text{CC_MAX}} = 2A \), \( V_{\text{TC}} = 3V \), \( V_{\text{IN_MAX}} = 10V \), the maximum ripple current is 1A. Select the input capacitors so that the temperature rise due to the ripple current does not exceed 10°C. Use ceramic capacitors with X5R or X7R dielectrics because of their low ESR and small temperature coefficients.

For most applications, use a 10µF capacitor. Besides, usually a small cap with at least 1uF (C1) from IN to GND is required to be put as much close as possible to the IC. For the input voltage is high to 10V, consider the spike when input insert, select the input capacitors (both the 22uF and 1uF) in 25V rating.

Selecting the Output Capacitor

The output capacitor C\(_{\text{SYS}}\) from the typical application circuit is in parallel with the SYS load. C\(_{\text{SYS}}\) absorbs the high-frequency switching ripple current and smooth the output voltage. Its impedance must be much less than that of the system load to ensure it properly absorbs the ripple current.

Use a ceramic capacitor because it has lower ESR and smaller size that allows us to ignore the ESR of the output capacitor. Thus, the output voltage ripple is given by:

\[
\Delta r = \frac{\Delta V_{\text{SYS}}}{V_{\text{SYS}}} = \frac{1 - \frac{V_{\text{SYS'}}}{V_{\text{SYS}}}}{8 \times C_{\text{SYS}} \times f_{s}^2 \times L} \% \quad (12)
\]

In order to guarantee the ±0.5% system voltage accuracy, the maximum output voltage ripple must not exceed 0.5% (e.g. 0.1%). The maximum output voltage ripple occurs at the minimum system voltage and the maximum input voltage.

The output capacitor can be calculated with Equation (13):

\[
C_{\text{SYS}} = \frac{1 - \frac{V_{\text{SYS_MIN}}}{V_{\text{IN}}}}{8 \times f_{s}^2 \times L \times \Delta r} \quad (13)
\]

When SYSFB pin is floating, output voltage ripple is the main concern to select the output capacitor (C\(_{\text{SYS}}\)). Refer to Table 5 for detail selection guide about the SYS capacitance selection under typical inputs.

### Table 5: SYS Capacitance Selection Guide

<table>
<thead>
<tr>
<th>SPEC</th>
<th>SYS Capacitance (C(_{\text{SYS}})) Selection</th>
</tr>
</thead>
<tbody>
<tr>
<td>5V</td>
<td></td>
</tr>
<tr>
<td></td>
<td>( V_{\text{IN}} )</td>
</tr>
<tr>
<td></td>
<td>( C_{\text{SYS}} = \frac{1 - \frac{V_{\text{SYS}}}{V_{\text{IN}}}}{8 \times f_{s}^2 \times L \times \Delta r} )</td>
</tr>
<tr>
<td></td>
<td>( C_{\text{SYS_MIN}} (\mu F) ) 7)</td>
</tr>
<tr>
<td></td>
<td>When SYSFB is Floating</td>
</tr>
<tr>
<td>5V</td>
<td>13.6</td>
</tr>
<tr>
<td>9V</td>
<td>13.3</td>
</tr>
<tr>
<td></td>
<td>( \Delta r=0.1% )</td>
</tr>
<tr>
<td></td>
<td>( L=1uH @V_{\text{IN}}=5V )</td>
</tr>
<tr>
<td></td>
<td>( L=2.2uH @V_{\text{IN}}=9V )</td>
</tr>
<tr>
<td></td>
<td>20</td>
</tr>
<tr>
<td></td>
<td>X5R,X7R</td>
</tr>
<tr>
<td></td>
<td>Application Required</td>
</tr>
</tbody>
</table>

**NOTE:**

7) For different voltage rating, capacitance will have different DC bias characteristic. Suppose a general condition, capacitance drops 40% under \( V_{\text{SYS}} = 4.4V \) under 10V rating, and 50% at 6.3V rating.
When SYSFB is programmed using external resistors, the control loop function is changed. A zero point is added around the cross over frequency of the DC gain, and this may result in the phase margin varied a lot, which may cause the unstable operation. To avoid this condition, a minimum capacitance requirement should be satisfied to make the pole point to compensate the zero point. This minimum capacitance is 20uF for a general application.

So, for the SYSFB programmed condition, the CSYS should be selected as max (CSYS_MIN, 20uF), CSYS_MIN is calculated from the formula of equation (13), as shown in Table 5. For better stability margin, select a 47uF ceramic capacitor with 6.3V and above voltage rating as the output capacitor in this case.

**Resistor Choose for NTC Sensor**

According to formulas in "Battery Temperature Monitor", the low and high temperature threshold ratio of MP2625B is 65% and 33.5%, so we can get the following:

\[
\frac{R_{T2}}{R_{T1}} = \frac{R_{NTC\_LOW}}{R_{NTC\_LOW}} = 65\% \\
\frac{R_{T1}}{R_{T2}} = \frac{R_{NTC\_HIGH}}{R_{NTC\_HIGH}} = 33.5\% 
\]

The two resistors R_{T1} and R_{T2} allow the high temperature limit and low temperature limit to be programmed independently. With this feature, MP2625B can fit most type of NTC thermistors and different operation temperature range requirement.

R_{T1} and R_{T2} can be calculated according to different type of NTC thermistor, operation temperature range and equation (14) and (15). We also provide calculation tool for convenient design.

For example, if the thermistor –NCP18XH103 is selected, and the application requires 0°C to 50°C operation temperature range. Check the datasheet of NCP18XH103 to get the following information:

At 0°C, the R_{NTC\_LOW}=27.219kΩ
At 50°C, the R_{NTC\_HIGH}=4.161kΩ

Using the calculation tool to calculate R_{T1}=7.11kΩ and R_{T2}= 25.61kΩ. Then select the most closing 1% accuracy resistance for R_{T1} and R_{T2}.

**PCB Layout Guideline**

It is important to pay special attention to the PCB layout to meet specified noise, efficiency and stability requirements. The following design considerations can improve circuit performance:

1) Route the power stage adjacent to their grounds. Aim to minimize the high-side switching node (SW, inductor), trace lengths in the high-current paths and the current sense resistor trace.

Keep the switching node short and away from all small control signals, especially the feedback network.

Place the input capacitor as close as possible to the IN and PGND pins.

Place the output inductor close to the IC and connect the output capacitor between the inductor and PGND of the IC.

2) For high-current applications, the balls for the power pads (IN, SW, SYS, BATT and PGND) should be connected to as much copper in the board as possible. This improves thermal performance because the board conducts heat away from the IC.

3) The PCB should have a ground plane connected directly to the return of all components through vias (two vias per capacitor for power-stage capacitors, one via per capacitor for small-signal components). It is also recommended to put vias inside the PGND pads for the IC, if possible. A star ground design approach is typically used to keep circuit block currents isolated (high-power/low-power small-signal) which reduces noise-coupling and ground-bounce issues. A single ground plane for this design gives good results. With this small layout and a single ground plane, there is no ground-bounce issue, and having the components segregated minimizes coupling between signals.
TYPICAL APPLICATION CIRCUITS

Figure 10: Typical Charge Application Circuit for 5V input with NTC Resistor Fixed

Table 6: The Key BOM of Figure 10

<table>
<thead>
<tr>
<th>Qty</th>
<th>Ref</th>
<th>Value</th>
<th>Description</th>
<th>Package</th>
<th>Manufacture</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>C1</td>
<td>1 μF</td>
<td>Ceramic Capacitor; 10V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C2</td>
<td>1 uF</td>
<td>Ceramic Capacitor; 6.3V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C3</td>
<td>100nF</td>
<td>Ceramic Capacitor; 16V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>CTMR</td>
<td>100nF</td>
<td>Ceramic Capacitor; 6.3V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>2</td>
<td>CSYS,CBATT</td>
<td>22μF</td>
<td>Ceramic Capacitor; 10V; X5R or X7R</td>
<td>1206</td>
<td>Any</td>
</tr>
<tr>
<td>2</td>
<td>RT1,RT2</td>
<td>10k</td>
<td>Film Resistor; 1%</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>L1</td>
<td>1.0μH</td>
<td>Inductor; 1.0μH; Low DCR; ISAT &gt; 2.8A</td>
<td>SMD</td>
<td>Any</td>
</tr>
</tbody>
</table>
TYPICAL APPLICATION CIRCUITS

Figure 11: Typical Charge Application Circuit for 9V Input and 1.5A Input Current Limit

Table 7: The Key BOM of Figure 11

<table>
<thead>
<tr>
<th>Qty</th>
<th>Ref</th>
<th>Value</th>
<th>Description</th>
<th>Package</th>
<th>Manufacture</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>C&lt;sub&gt;IN&lt;/sub&gt;</td>
<td>22μF</td>
<td>Ceramic Capacitor; 16V; X5R or X7R</td>
<td>1206</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C1</td>
<td>1μF</td>
<td>Ceramic Capacitor; 16V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C2</td>
<td>1μF</td>
<td>Ceramic Capacitor; 6.3V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C3</td>
<td>100nF</td>
<td>Ceramic Capacitor; 25V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C4</td>
<td>4.7μF</td>
<td>Ceramic Capacitor; 10V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>C&lt;sub&gt;TMR&lt;/sub&gt;</td>
<td>100nF</td>
<td>Ceramic Capacitor; 6.3V; X5R or X7R</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>2</td>
<td>C&lt;sub&gt;SYS&lt;/sub&gt;, C&lt;sub&gt;BATT&lt;/sub&gt;</td>
<td>22μF</td>
<td>Ceramic Capacitor; 10V; X5R or X7R</td>
<td>1206</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>R6</td>
<td>26.5k</td>
<td>Film Resistor; 1%</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>3</td>
<td>R&lt;sub&gt;T1&lt;/sub&gt;, R&lt;sub&gt;T2&lt;/sub&gt;, R7</td>
<td>10k</td>
<td>Film Resistor; 1%</td>
<td>0603</td>
<td>Any</td>
</tr>
<tr>
<td>1</td>
<td>L1</td>
<td>2.2μH</td>
<td>Inductor; 2.2uH; Low DCR; I&lt;sub&gt;SAT&lt;/sub&gt; &gt; 6A</td>
<td>SMD</td>
<td>Any</td>
</tr>
</tbody>
</table>
PACKAGE INFORMATION

QFN-20 (3mmX4mm)

TOP VIEW

BOTTOM VIEW

SIDE VIEW

NOTE:
1) ALL DIMENSIONS ARE IN MILLIMETERS
2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETERS MAX
4) JEDEC REFERENCE IS MO220.
5) DRAWING IS NOT TO SCALE

RECOMMENDED LAND PATTERN

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