DESCRIPTION

MP150 is a primary-side regulator that provides accurate constant voltage (CV) regulation without the opto-coupler, and supports Buck, Buck-boost, Boost and Flyback topologies. It has an integrated 500V MOSFET that simplifies the structure and reduces costs. These features help to make it a competitive candidate for offline low power applications, such as home appliances and standby power.

MP150 is a green-mode operation regulator. Both the peak current and switching frequency decrease as the load decreases to provide excellent efficiency performance at light load, thus improving the overall average efficiency.

The MP150 features various protections, including thermal shutdown (TSD), VCC under-voltage lockout (UVLO), over-load protection (OLP), short-circuit protection (SCP), and open loop protection.

MP150 is available in the TSOT23-5 and SOIC8 packages.

FEATURES

- Primary-side constant voltage (CV) control, supporting Buck, Buck-boost, Boost and Flyback topologies
- Integrated 500V/30Ω MOSFET
- < 150mW No-load power consumption
- Up to 2W output power
- Maximum DCM output current less than 120mA
- Maximum CCM output current less than 200mA
- Frequency foldback
- Maximum frequency limitation
- Peak current compression
- Internal high-voltage current source

APPLICATIONS

- Home Appliance, White Goods and Consumer Electronics
- Industrial Controls
- Standby Power

All MPS parts are lead-free, halogen free, and adhere to the RoHS directive. For MPS green status, please visit MPS website under Quality Assurance.

“MPS” and “The Future of Analog IC Technology” are Registered Trademarks of Monolithic Power Systems, Inc.
### ORDERING INFORMATION

<table>
<thead>
<tr>
<th>Part Number*</th>
<th>Package</th>
<th>Top Marking</th>
</tr>
</thead>
<tbody>
<tr>
<td>MP150GJ</td>
<td>TSOT23-5</td>
<td>ADG</td>
</tr>
<tr>
<td>MP150GS</td>
<td>SOIC8</td>
<td>MP150</td>
</tr>
</tbody>
</table>

* For Tape & Reel, add suffix –Z (e.g. MP150GJ–Z);  
* For Tape & Reel, add suffix –Z (e.g. MP150GS–Z);

### PACKAGE REFERENCE

**TOP VIEW**

<table>
<thead>
<tr>
<th>VCC</th>
<th>FB</th>
<th>SOURCE</th>
<th>DRAIN</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>2</td>
<td>3</td>
<td>5</td>
</tr>
</tbody>
</table>

**TSOT23-5**

<table>
<thead>
<tr>
<th>VCC</th>
<th>FB</th>
<th>SOURCE</th>
<th>N/C</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>2</td>
<td>3</td>
<td>5</td>
</tr>
</tbody>
</table>

**SOIC8**

<table>
<thead>
<tr>
<th>VCC</th>
<th>FB</th>
<th>SOURCE</th>
<th>DRAIN</th>
<th>N/C</th>
<th>N/C</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>2</td>
<td>3</td>
<td>4</td>
<td>5</td>
<td>6</td>
</tr>
</tbody>
</table>

### ABSOLUTE MAXIMUM RATINGS (1)

- Drain to SOURCE ......................... -0.7V to 500V  
- All Other Pins .............................. -0.7V to 6.5V  
- Continuous Power Dissipation \( T_A = +25^\circ C \) (2) \( 1W \)  
- TSOT23-5 .......................................... 1W  
- SOIC8.................................................. 1W  
- Junction Temperature ..................... 150°C  
- Lead Temperature ............................ 260°C  
- Storage Temperature ...................... -60°C to +150°C  
- ESD Capability Human Body Mode ........... 4.0kV  
- ESD Capability Machine Mode .............. 200V

### Recommended Operating Conditions (3)

- Operating Junction Temp. \( T_J \) -40°C to +125°C  
- Operating VCC range ....................... 5.3V to 5.6V

### Thermal Resistance (4) \( \theta_JA \theta_JC \)

- TSOT23-5 .................................................. 100 ...... 55 ... °C/W  
- SOIC8..................................................... 96 ...... 45 ... °C/W

**Notes:**

1) Exceeding these ratings may damage the device.  
2) The maximum allowable power dissipation is a function of the maximum junction temperature \( T_J(MAX) \), the junction-to-ambient thermal resistance \( \theta_JA \), and the ambient temperature \( T_A \). The maximum allowance continuous power dissipation at any ambient temperature is calculated by \( PD(MAX)=(TJ(MAX)-TA)/\theta_JA \). Exceeding the maximum allowance power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuit protects the device from permanent damage.  
3) The device is not guaranteed to function outside of its operating conditions.  
4) Measured on JESD51-7, 4-layer PCB.
**ELECTRICAL CHARACTERISTICS**

\( V_{CC} = 5.8V, \ T_A = 25^\circ C, \) unless otherwise noted.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Condition</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Units</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Start-up Current Source (Drain Pin)</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Internal regulator supply current</td>
<td>( I_{\text{regulator}} )</td>
<td>( V_{CC}=4V;V_{\text{Drain}}=100V )</td>
<td>2.5</td>
<td>3.5</td>
<td>4.5</td>
<td>mA</td>
</tr>
<tr>
<td>Drain pin leakage current</td>
<td>( I_{\text{Leak}} )</td>
<td>( V_{CC}=5.8V;V_{\text{Drain}}=400V )</td>
<td>10</td>
<td>12</td>
<td></td>
<td>μA</td>
</tr>
<tr>
<td>Breakdown voltage</td>
<td>( V_{BRDSS} )</td>
<td></td>
<td>500</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td><strong>Supply Voltage Management (VCC Pin)</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>VCC level (increasing) where the internal regulator stops</td>
<td>( V_{CCOFF} )</td>
<td></td>
<td>5.4</td>
<td>5.6</td>
<td>5.8</td>
<td>V</td>
</tr>
<tr>
<td>VCC level (decreasing) where the internal regulator turns on</td>
<td>( V_{CCON} )</td>
<td></td>
<td>5.1</td>
<td>5.3</td>
<td>5.6</td>
<td>V</td>
</tr>
<tr>
<td>VCC regulator on and off hysteresis</td>
<td></td>
<td></td>
<td>250</td>
<td></td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td>VCC level (decreasing) where the IC stops working</td>
<td>( V_{CC\text{stop}} )</td>
<td></td>
<td>3.4</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>VCC level (decreasing) where the protection phase ends</td>
<td>( V_{CC\text{pro}} )</td>
<td></td>
<td>2.4</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>Internal IC consumption</td>
<td>( I_{\text{CC}} )</td>
<td>( V_{CC}=5.8V, f_s=37kHz, D=40% )</td>
<td></td>
<td></td>
<td>430</td>
<td>μA</td>
</tr>
<tr>
<td>Internal IC consumption (no switching)</td>
<td>( I_{\text{CC}} )</td>
<td></td>
<td></td>
<td></td>
<td>300</td>
<td>μA</td>
</tr>
<tr>
<td>Internal IC Consumption, Latch off Phase</td>
<td>( I_{\text{CC\text{LATCH}}} )</td>
<td>( V_{CC}=5.3V )</td>
<td></td>
<td>16</td>
<td></td>
<td>μA</td>
</tr>
<tr>
<td><strong>Internal MOSFET (Drain Pin)</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Breakdown voltage</td>
<td>( V_{BRDSS} )</td>
<td></td>
<td>500</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>ON resistance</td>
<td>( R_{\text{on}} )</td>
<td></td>
<td></td>
<td></td>
<td>30</td>
<td>Ω</td>
</tr>
<tr>
<td><strong>Internal Current Sense</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Peak current limit</td>
<td>( I_{\text{Limit}} )</td>
<td></td>
<td>260</td>
<td>290</td>
<td>345</td>
<td>mA</td>
</tr>
<tr>
<td>Leading-edge blanking</td>
<td>( \tau_{\text{LEB1}} )</td>
<td></td>
<td>350</td>
<td></td>
<td></td>
<td>ns</td>
</tr>
<tr>
<td>SCP point</td>
<td>( I_{\text{SCP}} )</td>
<td></td>
<td>450</td>
<td></td>
<td></td>
<td>mA</td>
</tr>
<tr>
<td>Leading-edge blanking for SCP</td>
<td>( \tau_{\text{LEB2}} )</td>
<td></td>
<td>180</td>
<td></td>
<td></td>
<td>ns</td>
</tr>
<tr>
<td><strong>Feedback input (FB Pin)</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Minimum off time</td>
<td>( \tau_{\text{minoff}} )</td>
<td></td>
<td>15</td>
<td>18</td>
<td>21</td>
<td>μs</td>
</tr>
<tr>
<td>Primary MOSFET feedback turn-on threshold</td>
<td>( V_{FB} )</td>
<td></td>
<td>2.45</td>
<td>2.55</td>
<td>2.65</td>
<td>V</td>
</tr>
<tr>
<td>OLP feedback trigger threshold</td>
<td>( V_{FB_\text{OLP}} )</td>
<td></td>
<td></td>
<td>1.7</td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>OLP delay time</td>
<td>( \tau_{\text{OLP}} )</td>
<td>( f_s=37kHz )</td>
<td></td>
<td>170</td>
<td></td>
<td>ms</td>
</tr>
<tr>
<td>Open-loop detection</td>
<td>( V_{\text{OLD}} )</td>
<td></td>
<td>60</td>
<td></td>
<td></td>
<td>mV</td>
</tr>
<tr>
<td><strong>Thermal Shutdown</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Thermal shutdown threshold</td>
<td></td>
<td></td>
<td></td>
<td>150</td>
<td></td>
<td>°C</td>
</tr>
</tbody>
</table>
TYPICAL CHARACTERISTICS

Breakdown Voltage vs. Junction Temperature

On-State Resistance vs. Junction Temperature

Feedback Voltage vs. Junction Temperature

Minimum Off Time vs. Junction Temperature
TYPICAL PERFORMANCE CHARACTERISTICS

$V_{IN} = 265\text{VAC}, V_{OUT} = 5\text{V}, I_{OUT} = 200\text{mA}, L = 1\text{mH}, C_{OUT} = 100\mu\text{F}, T_A = +25^\circ\text{C},$ unless otherwise noted.

Start Up

Normal Operation

SCP

Open Loop Protection

FIGURE 1

FIGURE 2

FIGURE 3

FIGURE 4
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

\( V_{\text{IN}} = 230\text{VAC}, V_{\text{OUT}} = 5\text{V}, I_{\text{OUT}} = 200\text{mA}, L = 1\text{mH}, C_{\text{OUT}} = 100\mu\text{F}, T_{\text{A}} = +25^\circ\text{C}, \) unless otherwise noted.

Input Power Start Up

Input Power Shut Down

SCP Entry

SCP recovery

Open Loop Entry

Open Loop Recovery

Output Voltage Ripple

Load Transient
### PIN FUNCTIONS

<table>
<thead>
<tr>
<th>Pin # TSOT23-5</th>
<th>Pin # SOIC8</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>1</td>
<td>VCC</td>
<td>Control Circuit Power Supply.</td>
</tr>
<tr>
<td>2</td>
<td>2</td>
<td>FB</td>
<td>Regulator Feedback.</td>
</tr>
<tr>
<td>3,4</td>
<td>3,4</td>
<td>SOURCE</td>
<td>Internal Power MOSFET Source. Ground reference for VCC and FB pins.</td>
</tr>
<tr>
<td>5</td>
<td>7</td>
<td>DRAIN</td>
<td>Internal Power MOSFET Drain. High voltage current source input.</td>
</tr>
<tr>
<td>5,6,8</td>
<td>N/C</td>
<td></td>
<td>Not connected.</td>
</tr>
</tbody>
</table>
Figure 1: Functional Block Diagram
OPERATION

The MP150 is a green-mode-operation regulator. The peak current and the switching frequency both decrease as the load decreases to provide excellent efficiency at light load, and thus improve the overall average efficiency. The typical application diagram shows that the regulator operates using a minimal number of external components. It incorporates the following features:

Start-up and Under Voltage Lock-out

The internal high-voltage regulator supplies the IC from the Drain pin. The IC starts switching and the internal high voltage regulator turns off when the voltage on VCC reaches 5.6V. When the VCC voltage drops below 5.3V, the internal high voltage regulator turns on again to charge the external VCC capacitor. Use a capacitor in the several µF range stabilize the VCC voltage and this can lower the cost by decreasing the value of the capacitor.

When the voltage on VCC drops blow 3.4V, the IC stops, then the internal high-voltage regulator charges the VCC capacitor.

When faults occur, such as overload, short circuit, and over-heating, the IC stops working and an internal current source (16µA) discharges the VCC capacitor. Before the VCC voltage drops below 2.4V, the internal high-voltage regulator remains off and the VCC capacitor remains discharged.

Estimate the restart time after a fault as:

\[
t_{\text{restart}} = C_{\text{VCC}} \times \frac{V_{\text{CC}} - 2.4\text{V}}{16\text{µA}} + C_{\text{VCC}} \times \frac{5.6\text{V} - 2.4\text{V}}{3.5\text{mA}}
\]

Figure 2 shows the typical waveform with VCC under-voltage lockout.

Constant Voltage Operation

The MP150 is a fully-integrated regulator when used in a Buck solution as shown in the typical application on page 1.

The integrated MOSFET turns ON at the beginning of each cycle when the feedback voltage is below the reference voltage (2.5V), which indicates insufficient output voltage. The peak current limit determines the ON period. After the ON period elapses, the integrated MOSFET turns OFF. The freewheeling diode (D1) remains OFF until the inductor current charges the sampling capacitor (C3) voltage to the output voltage level. Then the sampling capacitor voltage changes with the output voltage. The sampling capacitor can sample and hold the output voltage to regulate the output voltage. The sampling capacitor voltage decreases after the inductor current drops below the output current. When the feedback voltage falls below the reference voltage (2.5V), a new switching cycle begins.

Figure 3 shows the detailed operation timing diagram under CCM.
Monitoring the sampling capacitor regulates the output voltage, as per the following equation:

\[ V_0 = 2.5V \times \frac{R_1 + R_2}{R_2} \]

**Frequency Foldback**

Under light load or no load conditions, the output drops very slowly, which increases the time for the MOSFET to turn ON again; i.e., frequency decreases as the load decreases. So the MP150 can maintain a high efficiency under light load condition by reducing the switching frequency automatically.

The switching frequency can be obtained as:

\[ f_s = \frac{(V_{in} - V_o)}{2L(I_{peak} - I_o)} \cdot \frac{V_o}{V_{in}}, \text{ for CCM} \]

\[ f_s = \frac{2(V_{in} - V_o)}{L I_{peak}^2} \cdot \frac{I_o V_o}{V_{in}}, \text{ for DCM} \]

At the same time, the peak current limit decreases from 290mA as the OFF time increases. In standby mode, the frequency and the peak current are both minimized, allowing for a small dummy load. As a result, the peak-current-compression function helps to reduce no-load consumption. Determine the peak current limitation from the following equation where \( \tau_{off} \) is the power module OFF time:

\[ I_{peak} = 290mA - \left(1mA / \mu s\right) \times (\tau_{off} - 18\mu s) \]

**Minimum OFF Time Limit**

The MP150 implements a minimum OFF time limit. During the normal operation, the minimum OFF time limit is 18\( \mu s \); during start up, the minimum OFF time limit gradually drops from 72\( \mu s \), to 36\( \mu s \), then to 18\( \mu s \) (see Figure 4). Each minimum OFF time has 128 switching cycles. This soft-start function allows for safe start-up.

**EA Compensation**

To improve load regulation, the MP150 implements an error amplifier (EA) compensation function (Figure 5). The MP150 samples the feedback voltage 6\( \mu s \) after the MOSFET turns off. EA compensation regulates the 2.5V reference voltage with the load, thus improving the power module regulation.

**RAMP Compensation**

An internal ramp compensation circuit precisely maintains the output voltage. An additional exponential voltage sinking source pulls down the feedback comparator’s reference voltage as shown in Figure 5. The ramp compensation is relative to the load conditions: Under full-load conditions, the compensation is \(\sim 1mV/\mu s\); With a

![](Figure3.png)  
**Figure 3: V\text{FB} vs V\text{OUT}**

![](Figure4.png)  
**Figure 4: t_{minoff} at Start-Up**

![](Figure5.png)  
**Figure 5: EA and Ramp Compensation**
decreasing load, the compensation increases exponentially.

**Over Load Protection (OLP)**
As the load increases, the peak current and the switching frequency increase with the load. When the switching frequency and peak current reaches their maximums, the output voltage will decrease if the load continues to increase. Then the FB voltage will drop below OLP threshold.

By continuously monitoring the FB voltage, the timer starts when the FB voltage drops below the 1.7V error flag threshold. Removing the error flag resets the timer. If the timer continues to completion at 170ms ($f_s = 37$kHz), OLP occurs. This timer duration avoids triggering OLP when the power supply starts up or enters a load transition phase, and therefore requires that the power supply start up in less than 170ms. A different switching frequency ($f_s$) changes the over-load protection delay time, as shown below:

$$\tau_{\text{Delay}} \approx 170 \text{ms} \times \frac{37 \text{kHz}}{f_s}$$

**Short-Circuit Protection (SCP)**
The MP150 shuts down when the peak current rises above 450mA as its short-circuit protection threshold. The power supply resumes operation after removing the fault.

**Thermal shutdown (TSD)**
To prevent from any lethal thermal damage, the MP150 shuts down switching when the inner temperature exceeds 150°C. During thermal shutdown (TSD), the VCC drops to 2.4V, and then the internal high voltage regulator recharges VCC.

**Open Loop Detection**
If the $V_{FB}$ drops below 60mV, the IC will stop working and begins a re-start cycle. The open-loop detection is blanked for 128 switching cycles during start-up.

**Leading-Edge Blanking**
An internal leading-edge blanking (LEB) unit between the current sense resistor inside the IC and the current comparator input avoids prematurely switching pulse termination due to the parasitic capacitance. During the blanking period, the current comparator is disabled and cannot turn off the external MOSFET. Figure 6 shows leading edge blanking.

![Figure 6: Leading-Edge Blanking](image-url)
## Application Information

### Table 1. Common Topologies Using MP150

<table>
<thead>
<tr>
<th>Topology</th>
<th>Circuit Schematic</th>
<th>Features</th>
</tr>
</thead>
<tbody>
<tr>
<td>High-Side Buck</td>
<td><img src="#" alt="Circuit Schematic" /></td>
<td>1. No-isolation, 2. Positive output, 3. Low cost, 4. Direct feedback</td>
</tr>
<tr>
<td>High-Side Buck-Boost</td>
<td><img src="#" alt="Circuit Schematic" /></td>
<td>1. No-isolation, 2. Negative output, 3. Low cost, 4. Direct feedback</td>
</tr>
<tr>
<td>Boost</td>
<td><img src="#" alt="Circuit Schematic" /></td>
<td>1. Isolation, 2. Positive output, 3. Low cost, 4. Indirect feedback</td>
</tr>
<tr>
<td>Flyback</td>
<td><img src="#" alt="Circuit Schematic" /></td>
<td>1. Isolation, 2. Positive output, 3. Low cost, 4. Indirect feedback</td>
</tr>
</tbody>
</table>
Topology Options
MP150 can be used in common topologies, such as Buck, Buck-Boost, Boost and Flyback. Please find the Table.1 for more information.

Component Selection

Input Capacitor
The input capacitor supplies the converter’s DC input voltage. Figure 7 shows the typical half-wave rectifier’s DC bus voltage waveform.

![Figure 7: Input Voltage Waveform](image)

When using the half-wave rectifier, set the input capacitor 3µF/W for the universal input condition. When using the full-wave rectifier, choose a smaller capacitor, but avoid a minimum DC voltage below 70V to avoid thermal shutdown.

Inductor
MP150 has a minimum OFF time limit that determines the maximum power output. The maximum power increases with the inductor value. Using a smaller inductor may cause the output to fail at full load, but a larger inductor results in a higher OLP load. The optimal inductor value is the smallest that can supply the rated power. The maximum power is:

$$P_{\text{omax}} = V_o \left( I_{\text{peak}} - \frac{V_o \tau_{\text{minoff}}}{2L} \right), \text{ for CCM}$$

$$P_{\text{omax}} = \frac{1}{2} L I_{\text{peak}}^2 \cdot \frac{1}{\tau_{\text{minoff}}}, \text{ for DCM}$$

To account for converter parameters—such as peak current limit and minimum OFF time—estimate the minimum inductor power ($P_{\text{min}}$) for the maximum power, and selecting an inductor with a $P_{\text{min}}$ value that exceeds the rated power.

Using output voltages 5V and 12V as examples, Figure 8 shows the curve for $P_{\text{min}}$ at 5V, and Figure 9 shows the curve for $P_{\text{min}}$ at 12V. ($I_{\text{peak}}=0.29A$, $\tau_{\text{minoff}}=18\mu s$).

Freewheeling Diode
Choose a diode with a maximum reverse voltage rating that exceeds the maximum input voltage, and a current rating that exceeds the output current.

The reverse recovery of the freewheeling diode can affect the efficiency and circuit operation. Select an ultra-fast diode, such as the EGC10JH for DCM and the UGC10JH for CCM.
Output Capacitor

The output capacitor maintains the DC output voltage. Estimate the output voltage ripple as:

\[ V_{CCM\_ripple} = \frac{\Delta i}{8f_s C_o} + \Delta i \cdot R_{ESR} \text{ for CCM} \]

\[ V_{DCM\_ripple} = \frac{l_o}{f_s C_o} \left( \frac{l_{pk} - l_o}{l_{pk}} \right)^2 + l_{pk} \cdot R_{ESR} \text{ for DCM} \]

Use ceramic, tantalum, or low-ESR electrolytic capacitors to lower the output voltage ripple.

Feedback Resistors

The resistor divider determines the output voltage. Choose appropriate R1 and R2 values to maintain the FB voltage at 2.5V. Avoid very large values for R2 (typical values between 5kΩ to 10kΩ).

Feedback Capacitor

The feedback capacitor provides a sample-and-hold function. Small capacitors result in poor regulation at light load condition, and large capacitors can impact circuit operation. Estimate the capacitor range as per the following equation:

\[ \frac{1}{2} \frac{v_o}{R_1 + R_2} \cdot C_o \leq C_{FB} \leq \frac{v_o}{R_1 + R_2} \cdot C_o \]

Choose an appropriate value given practical considerations.

Dummy Load

A dummy load maintains the load regulation. This ensures sufficient inductor energy to charge the sample-and-hold capacitor to detect the output voltage. Start with a 3mA dummy load and adjust as necessary.

Surge Performance

To obtain a good surge performance, select an appropriate input capacitor that meets different surge tests. Figure 10 shows the half-wave rectifier. Table 2 shows the required capacitance under normal conditions for different surge voltages.

![Figure 10: Half-Wave Rectifier](image_url)

<table>
<thead>
<tr>
<th>Surge voltage</th>
<th>500V</th>
<th>1000V</th>
<th>2000V</th>
</tr>
</thead>
<tbody>
<tr>
<td>C1</td>
<td>1μF</td>
<td>10μF</td>
<td>22μF</td>
</tr>
<tr>
<td>C2</td>
<td>1μF</td>
<td>4.7μF</td>
<td>10μF</td>
</tr>
</tbody>
</table>
Layout Guide

PCB layout is very important to achieve reliable operation, good EMI, and good thermal performance. Follow these guidelines to optimize performance.

1) Minimize the loop area formed by the input capacitor, IC part, freewheeling diode, inductor and output capacitor.

2) Place the power inductor far away from the input filter.

3) Add a capacitor in the several-hundred pF range between the FB and source pins, as close as to the IC as possible.

4) Connect the exposed pad with the Drain pin to a large copper area to improve thermal performance.

Design Example

Below is a design example following the application guidelines given the following specifications:

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Specification</th>
</tr>
</thead>
<tbody>
<tr>
<td>$V_{IN}$</td>
<td>85 to 265Vac</td>
</tr>
<tr>
<td>$V_{OUT}$</td>
<td>5V</td>
</tr>
<tr>
<td>$I_{OUT}$</td>
<td>200mA</td>
</tr>
</tbody>
</table>

Figure 12 shows the detailed application schematic. The Typical Performance Characteristics section lists typical performance and circuit waveforms. For more device applications, refer to the related Evaluation Board Datasheets.
TYPICAL APPLICATION CIRCUITS

Figure 11 shows a typical application example of a 5V, 200mA non-isolated power supply using MP150.

![Figure 11: Typical Application Example; 5V, 200mA](image_url)
FLOW CHART

Start

Internal High Voltage Regulator ON

Vcc>5.6V

Y

Vcc<5.3V

N

Vcc<3.4V

Y

Vcc Decrease to 2.4

Y

Shut Down Internal High Voltage Regulator

Soft Start

Monitor Vcc

N

Vcc Increase

Monitor VFB

N

VFB<2.5V

Y

VFB<1.7V

N

Turn ON the MOSFET

OLP=Logic High

Shut off the Switching Pulse

OTP, SCP or open loop Logic High?

N

OTP, SCP and open loop Monitor

N

Continuous Fault Monitor

6144 switching counter finished?

Y

N

Vcc<5.3V

Y

UVLO, OTP, SCP, OLP and Open Loop Protection are auto restar

Figure 12: Control Flow Chart
Figure 13: Signal Evolution Resulting from Faults
PACKAGE INFORMATION

TSOT23-5

TOP VIEW

RECOMMENDED LAND PATTERN

FRONT VIEW

SIDE VIEW

NOTE:

1) ALL DIMENSIONS ARE IN MILLIMETERS.
2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSION OR GATE BURR.
3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION.
4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.10 MILLIMETERS MAX.
5) DRAWING CONFORMS TO JEDEC MO-193, VARIATION AA.
6) DRAWING IS NOT TO SCALE.
NOTICE: The information in this document is subject to change without notice. Please contact MPS for current specifications. Users should warrant and guarantee that third party Intellectual Property rights are not infringed upon when integrating MPS products into any application. MPS will not assume any legal responsibility for any said applications.

MP150 – OFFLINE PRIMARY-SIDE REGULATOR

PACKAGE INFORMATION

SOIC8

TOP VIEW

0.189(4.80) 0.197(5.00)
8 5
PIN 1 ID
1 4

0.150(3.80) 0.228(5.80)
0.157(4.00) 0.244(6.20)

RECOMMENDED LAND PATTERN

SIDE VIEW

0.0075(0.19)
0.0099(0.25)

0.053(1.35)
0.069(1.75)
SEATING PLANE

0.004(0.10)
0.010(0.25)

SEE DETAIL “A”

FRONT VIEW

0.010(0.25) BSC
0.020(0.50)

0.013(0.33)
0.020(0.51)

0.050(1.27) BSC

0.050(1.27)

DETAL "A"

GAUGE PLANE 0.010(0.25) BSC
0.010(0.25) x 45°

DETAIL "A"

0°-8°

0.016(0.41)
0.050(1.27)

NOTE:

1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.
5) DRAWING CONFORMS TO JEDEC MS-012, VARIATION AA.
6) DRAWING IS NOT TO SCALE.

NOTICE: The information in this document is subject to change without notice. Please contact MPS for current specifications. Users should warrant and guarantee that third party Intellectual Property rights are not infringed upon when integrating MPS products into any application. MPS will not assume any legal responsibility for any said applications.