DESCRIPTION
The HFC0310 is a flyback controller with programmable fixed-frequency operation.

The controller uses peak current mode to provide excellent transient response and ease loop compensation. When the output power falls below a given level, the controller enters burst mode to lower the stand-by power consumption.

An external capacitor connected between the FSET pin and GND programs the HFC0310 switching frequency. Otherwise, the HFC0310 uses a frequency shaping function that greatly reduces the noise level, and reduces the cost of the EMI filter.

The HFC0310 provides various protections, such as thermal shutdown, $V_{CC}$ under-voltage lockout, over-load protection, over-voltage protection, and short-circuit protection.

The HFC0310 is available in a SOIC8 package.

FEATURES

- Programmable switching frequency up to 600kHz
- Frequency shaping
- Current-mode operation
- Very low start-up current
- Very low standby power consumption via active-burst mode
- Internal leading-edge blanking
- Built-in soft-start function
- Internal slope compensation
- External protection with recovery hysteresis on PRO pin
- Over-temperature protection
- $V_{CC}$ under-voltage lockout with hysteresis
- Over-voltage protection on VCC
- Time-based over-load protection
- Short-circuit protection

APPLICATIONS

- Power Meters
- Switching Mode Power Supplies
- AC/DC Adapters, Switching Chargers

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TYPICAL APPLICATION

[Diagram of HFC0310 Programmable Fixed-Frequency Flyback Controller]
### ORDERING INFORMATION

<table>
<thead>
<tr>
<th>Part Number*</th>
<th>Package</th>
<th>Top Marking</th>
</tr>
</thead>
<tbody>
<tr>
<td>HFC0310GS</td>
<td>SOIC8</td>
<td>See below</td>
</tr>
</tbody>
</table>

* For Tape & Reel, add suffix –Z (e.g. HFC0310GS–Z);

### TOP MARKING

HFC0310  
LLLLLLLL  
MPSYWW

MPS: MPS prefix  
Y: year code  
WW: week code  
HFC0310: part number  
LLLLLLLL: lot number

### PACKAGE REFERENCE

![Top View Diagram]
**ABSOLUTE MAXIMUM RATINGS** (1)

- **Vcc,Drv to GND** ........................................... –0.3V to 30 V
- **All Other Pins to GND** ............................. –0.3V to 7 V
- **Continuous Power Dissipation** \( T_A = +25^\circ C \) (2)

<table>
<thead>
<tr>
<th>Package</th>
<th>Power Dissipation</th>
</tr>
</thead>
<tbody>
<tr>
<td>SOIC8</td>
<td>1.3W</td>
</tr>
</tbody>
</table>

- **Junction Temperature** .................................. 150°C
- **Lead Temperature** ...................................... 260°C
- **Storage Temperature** ................................. –60°C to +150°C
- **ESD Capability Human Body Model** ................ 2.0kV
- **ESD Capability Machine Model** ...................... 200V

**Recommended Operation Conditions** (3)

- **Vcc to GND** ............................................. 11V to 20V

<table>
<thead>
<tr>
<th>Package</th>
<th>Thermal Resistance</th>
<th>( \theta_{JA} )</th>
<th>( \theta_{JC} )</th>
</tr>
</thead>
<tbody>
<tr>
<td>SOIC8</td>
<td>96</td>
<td>45</td>
<td>°C/W</td>
</tr>
</tbody>
</table>

**Notes:**

1. Exceeding these ratings may damage the device.
2. The maximum allowable power dissipation is a function of the maximum junction temperature \( T_J \) (MAX), the junction-to-ambient thermal resistance \( \theta_{JA} \), and the ambient temperature \( T_A \). The maximum allowable continuous power dissipation at any ambient temperature is calculated by \( P_D \) (MAX) = \( (T_J \) (MAX) – \( T_A \))/\( \theta_{JA} \). Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
3. The device is not guaranteed to function outside of its operating conditions.
4. Measured on JESD51-7, 4-layer PCB.
THE ELECTRICAL CHARACTERISTICS

VCC = 12V, TJ = -40°C to 105°C, min and max values are guaranteed by characterization, typical values are tested under 25°C, unless otherwise noted.

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Conditions</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Driving Signal (Pin Dv)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Sourcing Resistor</td>
<td>RH</td>
<td></td>
<td>17</td>
<td>25</td>
<td>34</td>
<td>Ω</td>
</tr>
<tr>
<td>Sinking Resistor</td>
<td>RL</td>
<td></td>
<td>9</td>
<td>13</td>
<td>17</td>
<td>Ω</td>
</tr>
<tr>
<td>Supply Voltage Management (Pin Vcc)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>VCC Upper Turn-On/Off Level</td>
<td>VCCH</td>
<td></td>
<td>10.8</td>
<td>12</td>
<td>13.2</td>
<td>V</td>
</tr>
<tr>
<td>VCC Lower Turn-On/Off Level</td>
<td>VCCL</td>
<td></td>
<td>8.5</td>
<td>9.5</td>
<td>10.5</td>
<td>V</td>
</tr>
<tr>
<td>Start-Up Current</td>
<td>IST</td>
<td>VCC=VCCH−0.5V, Before start up</td>
<td>12</td>
<td>20</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td>VCC OVP Level</td>
<td>VOVP</td>
<td></td>
<td>21</td>
<td>23</td>
<td>25</td>
<td>V</td>
</tr>
<tr>
<td>VCC Protection-Enabled Recharge Level</td>
<td>VCCR</td>
<td></td>
<td>5.8</td>
<td>6.4</td>
<td>7.0</td>
<td>V</td>
</tr>
<tr>
<td>Internal IC Consumption, Protection Phase</td>
<td>Ipro</td>
<td>VCC =6.0V</td>
<td>8</td>
<td>12</td>
<td></td>
<td>µA</td>
</tr>
<tr>
<td>Feedback Management (Pin FB)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Internal Pull-Up Resistor</td>
<td>RFB</td>
<td></td>
<td>12.5</td>
<td>14.5</td>
<td>17.0</td>
<td>kΩ</td>
</tr>
<tr>
<td>Internal Pull-Up Voltage</td>
<td>VUP</td>
<td></td>
<td>4.25</td>
<td>4.55</td>
<td>4.85</td>
<td>V</td>
</tr>
<tr>
<td>FB to Current-Set-Point Division Ratio</td>
<td>IDIV</td>
<td></td>
<td>3</td>
<td>3.5</td>
<td>4</td>
<td></td>
</tr>
<tr>
<td>Internal Soft-Start Time</td>
<td>TSS</td>
<td></td>
<td>3</td>
<td></td>
<td></td>
<td>ms</td>
</tr>
<tr>
<td>Falling FB Level Where the Regulator Enters Burst Mode</td>
<td>VBURL</td>
<td></td>
<td>0.43</td>
<td>0.5</td>
<td>0.57</td>
<td>V</td>
</tr>
<tr>
<td>Rising FB Level Where the Regulator Exits Burst Mode</td>
<td>VBURH</td>
<td></td>
<td>0.66</td>
<td>0.73</td>
<td>0.80</td>
<td>V</td>
</tr>
<tr>
<td>Over-Load Set Point</td>
<td>VOLP</td>
<td></td>
<td>3.5</td>
<td>3.75</td>
<td>4</td>
<td>V</td>
</tr>
<tr>
<td>Over-Load Delay Time</td>
<td>Tdelay</td>
<td>Fs=100kHz</td>
<td>82</td>
<td></td>
<td></td>
<td>ms</td>
</tr>
<tr>
<td>Timing Capacitor(FSET)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Maximum Voltage on the FSET Capacitor</td>
<td>VFSETmax</td>
<td></td>
<td>0.82</td>
<td>0.87</td>
<td>0.92</td>
<td>V</td>
</tr>
<tr>
<td>Source Current</td>
<td>IFSET</td>
<td></td>
<td>40</td>
<td>50</td>
<td>60</td>
<td>µA</td>
</tr>
<tr>
<td>FSET Capacitor Discharge Time (Active at drive turn on)</td>
<td>TDISCH</td>
<td></td>
<td>500</td>
<td></td>
<td></td>
<td>ns</td>
</tr>
<tr>
<td>Frequency Spectrum Shaping range, in percentage of Fs</td>
<td>RShaping</td>
<td></td>
<td>±3.5</td>
<td></td>
<td></td>
<td>%</td>
</tr>
<tr>
<td>Current Sampling Management (CS)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Leading-Edge Blanking for Current Sensor</td>
<td>TILEB1</td>
<td></td>
<td>180</td>
<td>280</td>
<td>400</td>
<td>ns</td>
</tr>
<tr>
<td>Leading-Edge Blanking for SCP</td>
<td>TILEB2</td>
<td></td>
<td>150</td>
<td>250</td>
<td>370</td>
<td>ns</td>
</tr>
<tr>
<td>Maximum Current Set-Point</td>
<td>VCS</td>
<td></td>
<td>0.90</td>
<td>0.97</td>
<td>1.03</td>
<td>V</td>
</tr>
<tr>
<td>Short-Circuit–Protection Set Point</td>
<td>VSCP</td>
<td></td>
<td>1.60</td>
<td>1.72</td>
<td>1.84</td>
<td>V</td>
</tr>
<tr>
<td>Internal-Slope–Compensation Ramp</td>
<td>SRamp</td>
<td>fs=100kHz</td>
<td>27</td>
<td>42</td>
<td>57</td>
<td>mV/µs</td>
</tr>
<tr>
<td>Protection Management (PRO)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Protection Voltage</td>
<td>VPRO</td>
<td></td>
<td>3.1</td>
<td>3.3</td>
<td>3.5</td>
<td>V</td>
</tr>
<tr>
<td>Protection Hysteresis</td>
<td>VHY</td>
<td></td>
<td>0.2</td>
<td></td>
<td></td>
<td>V</td>
</tr>
<tr>
<td>Thermal Shutdown</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Thermal shutdown threshold</td>
<td>(5)</td>
<td></td>
<td>150</td>
<td></td>
<td></td>
<td>°C</td>
</tr>
<tr>
<td>Thermal shutdown recovery hysteresis</td>
<td>(5)</td>
<td></td>
<td>40</td>
<td></td>
<td></td>
<td>°C</td>
</tr>
</tbody>
</table>

Notes:
5) This parameter is guaranteed by design.
## PIN FUNCTIONS

<table>
<thead>
<tr>
<th>Package Pin #</th>
<th>Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>FSET</td>
<td>Switching Frequency Set. Connect a capacitor to GND to set the switching frequency up to 600kHz.</td>
</tr>
<tr>
<td>2</td>
<td>Vcc</td>
<td>IC Power Supply. Connect to a 47μF bulky capacitor and a 0.1μF ceramic capacitor for most applications.</td>
</tr>
<tr>
<td>3</td>
<td>NC</td>
<td>Not Connected.</td>
</tr>
<tr>
<td>4</td>
<td>Drv</td>
<td>Drive Signal Output.</td>
</tr>
<tr>
<td>5</td>
<td>CS</td>
<td>Primary Current Sense.</td>
</tr>
<tr>
<td>6</td>
<td>PRO</td>
<td>Pull up PRO to shut down the IC with hysteresis.</td>
</tr>
<tr>
<td>7</td>
<td>GND</td>
<td>Ground.</td>
</tr>
<tr>
<td>8</td>
<td>FB</td>
<td>Feedback. The output voltage from the external compensation circuit is fed into this pin. This pin and the current sense signal from Source determines the PWM duty cycle. Burst mode operation and Over Load Protection are also detected on it.</td>
</tr>
</tbody>
</table>
TYPICAL CHARACTERISTICS

- **VCC Upper Level at which the IC switch on vs. Temperature**
- **VCC Lower Level at which the IC switch off vs. Temperature**
- **VCC OVP Level vs. Temperature**
- **Pin FB to Current Set point Division Ratio vs. Temperature**
- **FB Decreasing Level at which the Regulator enter the Burst Mode vs. Temperature**
- **VCC Re-Charge Level at which the protections occurs vs. Temperature**
- **FB Increasing Level at which the Regulator leave the Burst Mode vs. Temperature**
- **Over Load Set Point vs. Temperature**
- **Over Load Delay Time vs. Temperature**
TYPICAL CHARACTERISTICS (continued)

Maximum Voltage On FSET Capacitor vs. Temperature

Maximum Current Set-point vs. Temperature

Short Circuit Protection Set Point vs. Temperature Chart

Internal Slope Compensation Ramp vs. Temperature Chart

Protection Voltage vs. Temperature

Maximum Voltage On FSET Capacitor (V)

Maximum Current Set-point (V)

Short Circuit Protection Set Point (V)

Internal Slope Compensation Ramp (μV/μA)

Protection Voltage (V)

-50 -30 -10 10 30 50 70 90 110 130

-50 -30 -10 10 30 50 70 90 110 130

-50 -30 -10 10 30 50 70 90 110 130

-50 -30 -10 10 30 50 70 90 110 130

-50 -30 -10 10 30 50 70 90 110 130

-50 -30 -10 10 30 50 70 90 110 130

-50 -30 -10 10 30 50 70 90 110 130
TYPICAL PERFORMANCE CHARACTERISTICS

$V_{IN}=230\text{VAC}$, $V_{OUT1}=12\text{V}/0.8\text{A}$, $V_{OUT2}=8\text{V}/0.2\text{A}$, $V_{OUT3}=8\text{V}/0.05\text{A}$, $T_{A}=+25^\circ\text{C}$, unless otherwise noted.
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

$V_{IN}=230\text{VAC}$, $V_{OUT1}=12\text{V/0.8A}$, $V_{OUT2}=8\text{V/0.2A}$, $V_{OUT3}=8\text{V/0.05A}$, $T_a=+25^\circ\text{C}$, unless otherwise noted.

SCP Entry  
Full Load

SCP Recovery  
Full Load

SCP Startup  
Full Load

OVP Entry  
Full Load

OVP Recovery  
Full Load

OVP Startup  
Full Load

OCP Protection  
Full Load

Pro Protection  
$V_{IN}=85\text{V}, 160\text{V Protection, Full Load}$
BLOCK DIAGRAM

Figure 1: Functional Block Diagram
OPERATION

The HFC0310 incorporates all the necessary features to build a reliable switch-mode power supply. Its high level of integration requires very few external components. It has burst-mode operation to minimize the stand-by power consumption at light load. Protection features—such as auto-recovery for over-load protection (OLP), short-circuit protection (SCP), over-voltage protection (OVP), or thermal shutdown (TSD) for over-temperature protection (OTP)—contribute to a safer converter design without increasing circuit complexity.

PWM Operation

The HFC0310 is a fully integrated converter with adjustable-frequency peak-current-mode control PWM switching regulators. The output voltage is measured at FB through a resistive voltage divider, amplifier, and optocoupler. The voltage at the FB pin is compared to the internally measured switch current to control the output voltage. The integrated MOSFET turns on at the beginning of each clock cycle. The current in the inductor increases until it reaches the value set by the FB voltage, and then the integrated MOSFET turns off.

Start-Up and Vcc UVLO

During start-up, the IC consumption is IST, and the current supplied through the start-up resistor charges the VCC capacitor.

The IC starts switching and the operation current increases when VCC reaches VCH. At this point, the transformer’s auxiliary winding powers the IC. When VCC falls below VCL, the regulator stops switching and the current through the start up resistor charges the VCC capacitor again.

The lower threshold of VCC under-voltage lock-out (UVLO) decreases from VCL to VCR when fault conditions occur, such as OLP, OVP, and OTP.

Soft-Start

To reduce stress on the primary MOSFET and the secondary diode during start-up and to smoothly establish the output voltage, the HFC0310 has an internal soft-start circuit that gradually increases the primary current sense threshold, which determines the MOSFET peak current during start-up. The pulse-width of the power switching device progressively increases to establish optimal operating conditions until the feedback control loop takes charge.
Switching Frequency
The capacitor between the FSET pin and GND sets the switching frequency of the HFC0310. Estimate the oscillator frequency as per the equation below:

\[ f_S = \frac{1}{t_{\text{DISCH}} + C \times \frac{V_{\text{SET \, max}}}{V_{\text{SET}}}} \text{Hz} \]

Over Voltage Protection
Monitoring the VCC pin with about 20µs delay time allows the HFC0310 to enter OVP during an over-voltage condition; when VCC goes above V_OVP, HFC0310 will resume operation after the fault disappears.

Over-Current Protection
The HFC0310 continuously monitors the FB pin. When FB pulls up to V_{OLP}, if after a 8192 switching cycle delay the fault signal is still present, the HFC0310 shuts down as soon as the power supply undergoes an overload. When the fault disappears, the power supply resumes operation.

Short-Circuit Protection
By monitoring the CS pin, the HFC0310 shuts down when the voltage rises higher than V_{SCP}, to indicate a short circuit. The HFC0310 enters a safe low-power mode that prevents any lethal thermal or stress damage. As soon as the fault disappears, the power supply resumes operation.

Thermal Shutdown
When the temperature of the IC exceeds thermal shutdown threshold, the OTP is activated and it will resume operation when junction temperature drops to thermal shutdown recovery point.

Burst Operation
To minimize stand-by power consumption, the HFC0310 implements burst mode at no load or light load. As the load decreases, the FB voltage decreases. The IC stops switching when the FB voltage drops below the lower threshold, V_{BRUL}. Then the output voltage drops at a rate dependent on the load. This causes the FB voltage to rise again due to the negative feedback control loop. Once the FB voltage exceeds the upper threshold, V_{BRUH}, the switching pulse resumes. The FB voltage then decreases and the whole process repeats. Burst-mode operation alternately enables and disables the switching pulse of the MOSFET. Hence switching loss at no load or light load conditions is greatly reduced.

Figure 4 shows the signals generated by burst-mode operation.

PRO Pin
The PRO pin provides extra protection against abnormal conditions. Use the PRO pin for input OVP and/or other protections. If the PRO pin voltage exceeds V_PRO, the IC shuts down. As soon as the fault disappears, the power supply resumes operation.

Leading-Edge Blanking (LEB)
In normal operation, a resistor is placed between the MOSFET Source and Ground to senses the primary peak current. The FB voltage sets the turn-off threshold of the MOSFET, V_{SENSE}=V_{FB}/IDIV. HFC0310 turns off the MOSFET when the sensing resistor voltage rises to V_{SENSE}.

During start-up and over-load condition, the maximum primary peak current threshold is internally limited to Vcs to avoid excessive output power and lower the switch voltage stress.
In order to avoid turning off the MOSFET by mis-triggered spikes shortly after the switch turns on, the IC implements a leading-edge blanking period. During blanking time, any trigger signal on the source pin is blocked. Figure 5 shows the primary-current–sense waveform and the leading-edge blanking.

![Figure 5: Leading-Edge Blanking](image)

**Design Example**
The following is a design example using the application guidelines for the given specifications:

<table>
<thead>
<tr>
<th>V_IN</th>
<th>85V to 420V</th>
</tr>
</thead>
<tbody>
<tr>
<td>V_OUT_1</td>
<td>12V</td>
</tr>
<tr>
<td>V_OUT_2</td>
<td>8V</td>
</tr>
<tr>
<td>V_OUT_3</td>
<td>8V</td>
</tr>
<tr>
<td>f_SW</td>
<td>100kHz</td>
</tr>
</tbody>
</table>

The detailed application schematic is shown in Figure 6. The typical performance and circuit waveforms have been shown in the Typical Performance Characteristics section. For more possible applications of this device, please refer to the related Evaluation Board datasheets.
TYPICAL APPLICATION CIRCUITS

Figure 6: Typical Application Schematic
FLOW CHART

Start

Charge Voc capacitor through start up resistors

Vcc > VccOH

Y

N

Soft Start

Monitor Voc

Auxiliary winding supply the IC

Vcc < VccOL

Y

N

Shut off the building

TSD logic high

Y

N

Pin PRO Monitor

Logic Low

N

Thermal Monitor

Y

Vcc Decrease to Vref

N

8122 counter ?

OLP logic high

Y

N

Continuous Fault Monitor

Vcc < Vref

N

Fixed switching frequency operation

Vref > Vsurh

Y

N

Switch Off

Vref < Vsurh
SIGNAL EVOLUTION IN THE PRESENCE OF FAULTS
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HFC0310 – PROGRAMMABLE FIXED-FREQUENCY FLYBACK CONTROLLER

PACKAGE INFORMATION

SOIC8

TOP VIEW

RECOMMENDED LAND PATTERN

FRONT VIEW

SIDE VIEW

NOTE:

1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.
5) DRAWING CONFORMS TO JEDEC MS-012, VARIATION AA.
6) DRAWING IS NOT TO SCALE.

DETAIL "A"

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